

Pen-Raised Quail Hunt
6.111 Final Project

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ABSTRACT

Duck Hunt is among the most famous of the classic console games. This project sought to recreate the Nintendo classic using the MIT 6.111 FPGA Labkit and a video camera instead of the memorable Nintendo gun. To aim, the player wears a glove with two different color regions which the video camera uses to track the player's in-game gun-sight and to shoot. Tracking is accomplished using video processing, and the color regions that the image processing is sensitive to is completely customizable by the user at run-time using the labkit's buttons as inputs and VFD display as feedback. The game engine renders a recreation of the classic Duck Hunt playing field, but unlike the original, it is upsampled to be displayed on a 1024x768 XVGA display. The final implemented game met all given specifications and proved to be a faithful recreation of the original classic except for certain elements which added a socio-political commentary that expressed the views of the authors on the current state of American politics.

Rev 1: Added Verilog Appendix

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OVERVIEW

System Inputs and Outputs

The primary input to the game which moves the player's gun sight is a glove with two different color regions. The player moves his hand in front of a video camera, and the video camera tracks the glove's color regions to create the player's in-game coordinates. Although the game is played at a low resolution to maintain the original Nintendo look, it is rendered as a 1024x768 XVGA signal. Push-buttons are used to select between game-play and camera calibration.

The player can use switches and another set of buttons on the labkit to customize the input camera filters to be sensitive to any two color regions he desires. This customization allows the user to fine-tune the default color filter settings to suit his particular lighting conditions and camera. This also allows the player to use any input device he can conceive of that has two distinct color regions; the ambitious enthusiast is encouraged to tweak the system for maximum accuracy. Filter settings are displayed on the labkit's VFD display. The filtered video signal is displayed onto the XVGA output through an in-game calibration screen.

Figure 1 is a summary of the system's input and output.

Video Decoding and Processing Overview

The primary objective of the Video Decoding and Processing blocks is to generate player coordinates and a fire control signal for the game engine. This is done by looking at the input NTSC video stream from the camera and filtering for regions of user-defined color. The coordinates of pixels that pass the filters are added to a center-of-mass calculator which simply averages all passing pixels' coordinates. These center-of-mass averages, updated once per NTSC frame, are fed into the game engine as the player's coordinates.

The secondary objective of the Video Decoding and Processing blocks is to display the results of the color filtering and the center-of-mass calculation to the user. This proved to be one of the initial significant design challenges. The main problem associated with displaying the video feed to the user is that the user's display is a XVGA signal clocked at 60mhz while the video feed operates on a 27mhz clock. Because these two different clock rates are not simple multiples of each other, they had to be operated asynchronously. A ZBT RAM module was used as a frame buffer to solve this problem, but because the ZBT module is a single-port memory chip, a multiplexing scheme had to be designed to allow proper operation. A summary of the Video Decoding and Processing blocks is shown in Figure 2.

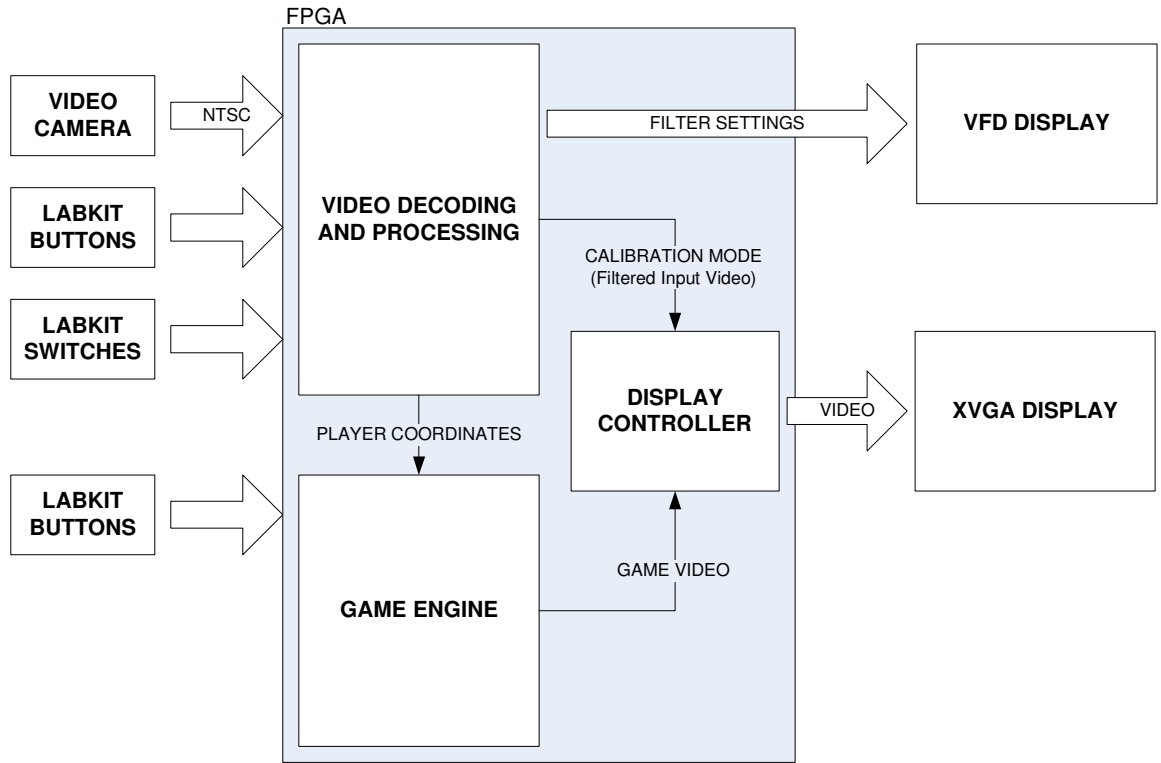


Figure 1: Summary of System Inputs and Outputs

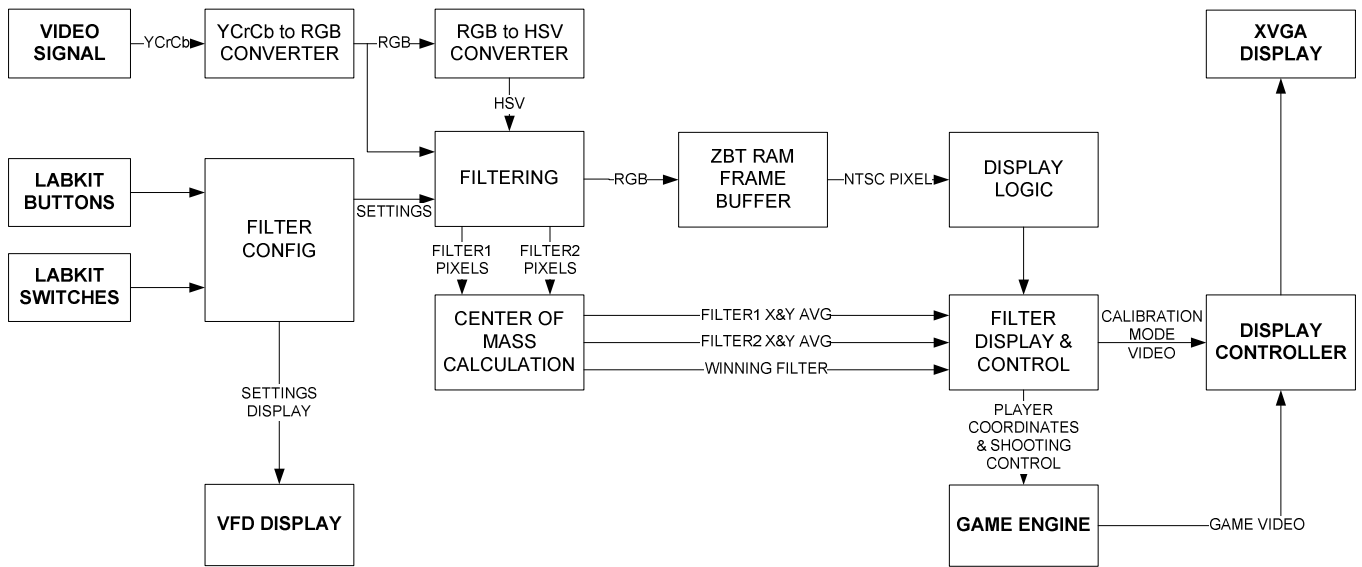


Figure 2: Summary of Video Decoding and Processing Blocks Logic

Filtering Overview

The system actually needs to look for two separate regions of color: one corresponding to a “track” hand gesture and one corresponding to a “fire” hand gesture. For example, in the project demonstration, a glove was provided that had a yellow-colored square on the part of the hand that was visible to the camera when the hand was open; this region corresponded to the game engine merely tracking the user’s hand and moving his gun-sight appropriately. When the user closed his hand into a fist, a blue-colored square was attached to the part of the hand that was visible to the camera; this region corresponded to the game engine shooting at the flying duck. The Video Decoding and Processing blocks actually contain *two* color-region filters which can be independently adjusted to be sensitive to different regions of color. The filter which had the most passing pixels (and therefore tells the logic which coordinates to send to the game engine and whether or not the user is firing or just tracking) is representing using the WINNING_FILTER line.

The goal of the color filtering is to be able to specify regions of colors that are of interest to the user; if the region on the player’s glove corresponding to tracking is a yellow-colored square, we want to be able to tweak the color filter to be sensitive only to regions of yellow. We are not interested in the intensity of the yellow since this can change due to different lighting conditions or camera. What is of interest to us is to specify the filter to be sensitive to all shades of “yellow.” The problem now at hand is which color-space would give the solution to this problem the most intuitive customization. Three color-spaces were considered: RGB, YCrCb, and HSV.

RGB color space is defined as the “intensity” of the red, green, and blue elements of a pixel. This makes the RGB color space poorly-suited to easily select a region of a particular “color”; to ignore intensity and concentrate on the desired color, one must look at the *differential* across the three channels. The RGB color space was quickly disregarded as a potential color space to do the filtering work.

The next color space considered was the camera’s native YCrCb color space. This color space was attractive because intensity is defined in the Y (or luminance) channel and “color” is defined using the Cr and Cb (or Chrominance-red and Chrominance-blue, respectively) channels. If intensity is not the main parameter, it could easily be ignored. However, specifying a particular “color” is initially non-intuitive because the “color” is a function of two variables – Cr and Cb. One of the design objectives of the project was to allow very simple and intuitive control over the color region of interest, and so this color space was disregarded on the grounds of it being too difficult to specify a particular color region.

The final color space that was considered was the HSV, or Hue-Saturation-Value color space. In HSV, the value of the Hue channel specifies a color, the value of the Saturation channel specifies how much “white” there is in the pixel, and the value of the Value channel specifies how much “black” there is in the pixel (equal Saturation and Value in a color shifts the color towards neutral-gray). Thus, specifying a region

of “color” can be easily accomplished by specifying a range of Hues and ignoring the Saturation and Value channels. A visualization of the HSV color space is shown in Figures 3a and 3b.

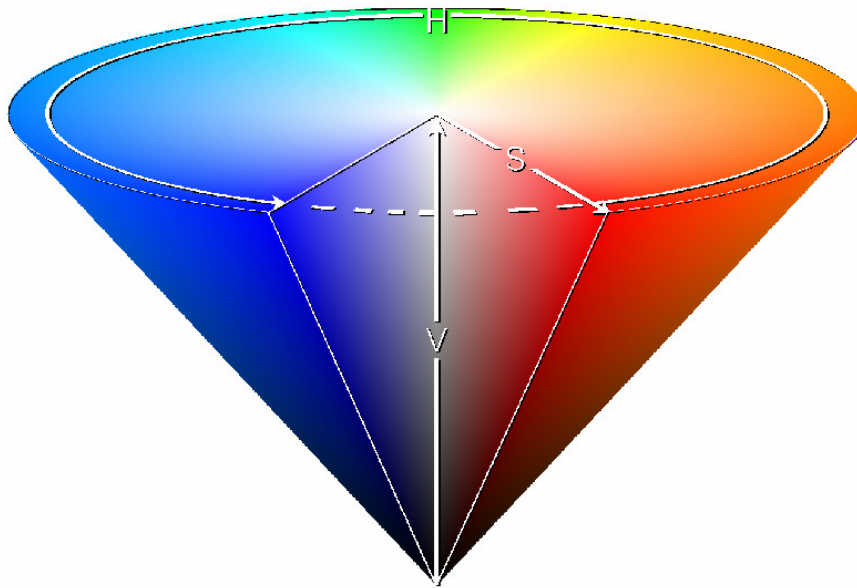


Figure 3a: 3D Conical Representation of the HSV Color Space

Graphic Source: Wikipedia (http://en.wikipedia.org/wiki/HSV_color_space)

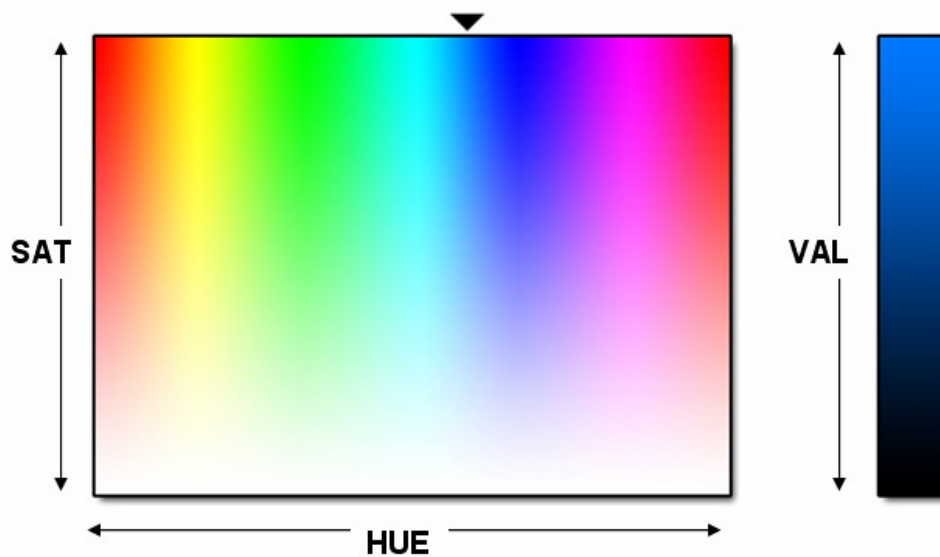


Figure 3b: 2D Representation of the HSV Color Space

Due to the simplicity of specifying a particular “color” in the HSV color space (a color is the function of just one parameter instead of two, as in YCrCb), it was

decided that filtering would be done with a transformation into the HSV coordinate space.

Filtering was originally implemented by just specifying a low cut-off for Hue and a high cut-off for Hue. However, during testing, it was discovered that whiter pixels tended to transform into the full spectrum of hues; a pure white pixel technically has an undefined hue, but as soon as a little amount of noise is added, the Hue value of the HSV transformation can easily become almost any hue. In other words, it was found that when a white-ish region was picked up by the camera, it tended to transform into some pixels in every hue-pass region, creating noise for the center-of-mass calculator. It was decided that a way to get rid of this noise was to filter out low saturation pixels, in essence adding a Saturation-Pass region to the Hue-Pass region.

Thus, each of the system's Color Filters is in fact a customizable Hue-Pass Filter and a customizable Sat-Pass Filter. For a pixel to pass one of the Color Filters, it must pass both the Hue-Pass Filter (if enabled) AND the Saturation-Pass Filter (if enabled). It was found that near-black pixels did not create a significant amount of the same problem, and so no filtering is done on the Value channel (although future revisions of the project should include Value-Pass filter option). Each color filter is therefore parameterized by four values: a Hue-Pass Low Mark, a Hue-Pass High Mark, a Saturation-Pass Low Mark, and a Saturation-Pass High Mark. The system has two independent color filters, so for the user to be able to completely customize the filters, the Filter Configuration module must be capable of adjusting eight independent parameters. Figure 4 is a visualization of one of the system's two color filters, adjusted to pass "blue" pixels.

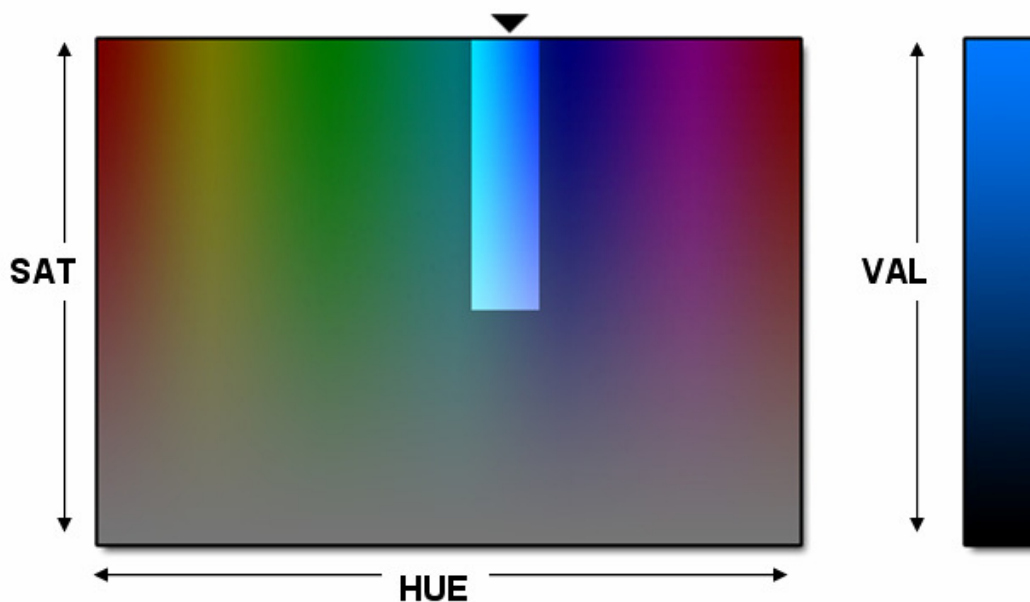


Figure 4: A Color Filter Adjusted To Pass “Blue” Pixels

AI Overview

The AI Module is responsible for controlling all actions of the AI player. This module receives state information from the Game Module, such as the duck’s position and heading, and generates a decision on what action to take next. This decision will include both considering where to move its crosshair as well as whether or not to fire. The AI has multiple difficulty settings, which affect its behavior. On higher difficulty settings, it tracks the duck extremely well and fires accurately, while on lower difficulty settings, it tracks the duck poorly and misses frequently. This is implemented by adding a small error term (the magnitude of which varies by difficulty) to the AI’s perception of the duck’s current and future location. By making the AI purposely miss and track without perfect precision, a more realistic, human-like opponent is simulated.

Game Module Overview

The Game Module is responsible for storing and updating all game state information, enforcing all game rules, and providing display data to the display controller. The game state, which includes position and status of all game objects, is updated once per frame of video. For example, at the beginning of every new frame of video, the duck is moved to a new position, any attempted shots are processed, etc. Game rules include game time limit, shots per round, scoring, etc. This module does not generate the display data internally, but rather sends state information to the various graphics modules, which return appropriate graphics information.

Graphics Pipeline Overview

The graphics pipeline is not a single module, but rather a collection of related graphics modules organized into a pipeline. Each module stores the bitmaps related to a single game object, for example the duck or a crosshair. The Game Module then sends an hCount, vCount pair to each module, which represents the coordinate of a pixel on screen. If a given module represents an object which has a pixel at that location, it will output the RGB value of that pixel, and assert a hasPixel signal which indicates that it has color information; otherwise, it will output nothing, and not assert hasPixel. All modules compute the RGB color value of a given pixel in parallel, and the module highest in the pipeline that asserts hasPixel will have its color information displayed on screen, for the pixel in question.

PROJECT DESCRIPTION

Overview of Video Decoding and Processing Modules

Authorship: Daniel Lopuch

The overall architecture of the Video Decoding and Processing Modules is based loosely off of the “ZBT RAM Example” sample code from the 6.111 Fall 2005 website. Although this sample file proved to be a good reference for code on decoding the raw NTSC data stream into YCrCb pixels and for interfacing with the ZBT memory, the shortcomings of the sample file were made painfully obvious when the jump from black and white to color was made. An overview of this project’s video decoding and processing modules is illustrated in Figure 5.

The main issues that had to be resolved involved multiplexing the writing and reading of the pixel data. The NTSC decoder (which generates the data to be written) operates at 27mhz while the XVGA display (which reads the data) operates at 60mhz. These different clock rates result in asynchronous read and write requests. The “ZBT RAM Example” source took advantage of the fact that when interested only at the black and white data, it is possible to store 4 pixels in one word of ZBT RAM. The XVGA display would read from the ZBT RAM once every four pixels and let the NTSC-to-ZBT module write during the off-time.

When color information is introduced, it is no longer possible to store four pixels in one word of ZBT RAM. ZBT RAM words are 36-bits long. It is possible to store *two* color pixels per ZBT RAM word if one decides to truncate the two least significant bits from the 8-bit RGB channels, but this approach was not considered because maximum precision was desired for the color filtering and this approach would lose some color depth. It was therefore decided that the system would store one pixel per word, and the problem became how to allow the NTSC processing module to write to the frame buffer when the XVGA display was reading from it during every new pixel.

The solution to this problem was developed when the realization was made that the XVGA display of the NTSC video stream is for the player’s feedback only and the true accuracy of the picture really does not matter as long as the player doesn’t perceive any imperfections. It was decided to allow the NTSC processing module to write to the ZBT buffer whenever it had new data. Whenever it did not have any new data, it would allow the display module to read whatever pixels it needed. During those pixels where the NTSC processing module wrote new data, the display module would read out junk data because it was overwritten by a write-request. This resulted in “snow” noise that would drift up along the displayed XVGA signal.

The “snow” was fixed by using a simple interpolator. The junk pixels could be predicted by delaying the write-enable line from the NTSC processor. Whenever the display module would read a junk pixel, it would simply override that pixel with the

previous clock cycle's pixel. This is accomplished with the "Frame Buffer Write-Conflict Interpolator" in Figure 5. This proved to be a very effective solution – although the displayed picture was technically not the exact NTSC picture, the interpolation was on such a fine scale that the error was unperceivable.

Another significant difference between the "ZBT RAM Example" source and the developed system was in the number of pixels stored in RAM. Because the example source used four pixels per word, it was able to store the entire 1024x768 XVGA frame in RAM; the display module simply read off and displayed each pixel. However, the 512k ZBT RAM chip is not big enough to do the same storing one pixel per word. Instead, only the 720x480 NTSC frame is stored in the RAM, and any XVGA pixels outside of this range are simply masked to black after the display module receives the pixel from the RAM (see "NTSC Mask" in Figure 5).

After the pixel information is retrieved from RAM and passes through the Frame Buffer Write-Conflict Interpolator and the NTSC Mask, it is passed to a Filter Display Control module. This module has two purposes. The first is to read the center-of-mass averages from the Filtering module and make a visualization of them across the entire NTSC frame, and the second purpose is to determine which center-of-mass should be used by the game engine and to scale it to the appropriate coordinate space. The center-of-mass visualization is just vertical and horizontal lines that form crosshairs onto the center-of-masses of each of the two color filters in the NTSC video frame. The crosshairs change color depending on which color filter passed the most amount of pixels (logically corresponding to which of the two center-of-mass averages the game engine should use as the player coordinates and whether the game should interpret the position as a tracking position or as a firing position). The filter that has the most passing pixels is represented using the winningFilter control line. Finally, the Filter Display and Control scales the winning average from the 720x480 coordinate space into the 1024x768 coordinate space used by the Game Engine and gives these coordinates to the Game Engine.

The VRAM Display Module simply uses hcount and vcount to generate an address with appropriate NTSC clipping and then passes the RAM data onto the other logic, and the Filter Display and Control is simply a collection of combinational logic that overrides the current pixel with an appropriately colored line if the pixel lines on a centroid's x or y coordinate. Because the simplicity of these modules, they will not be discussed in any detail.

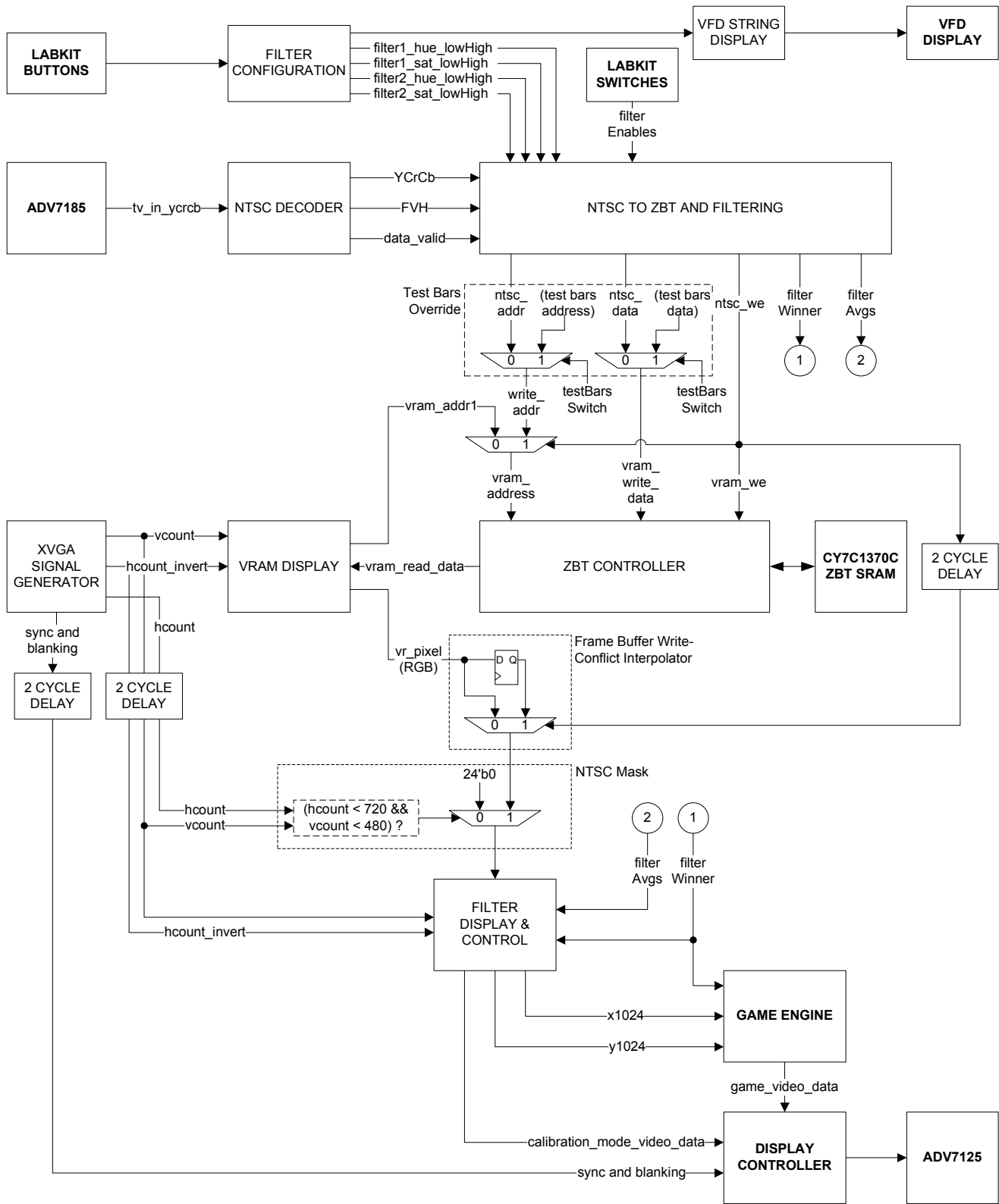


Figure 5: Overview of Video Decoding and Processing Verilog Modules

Filter Configuration Module

Authorship: Daniel Lopuch

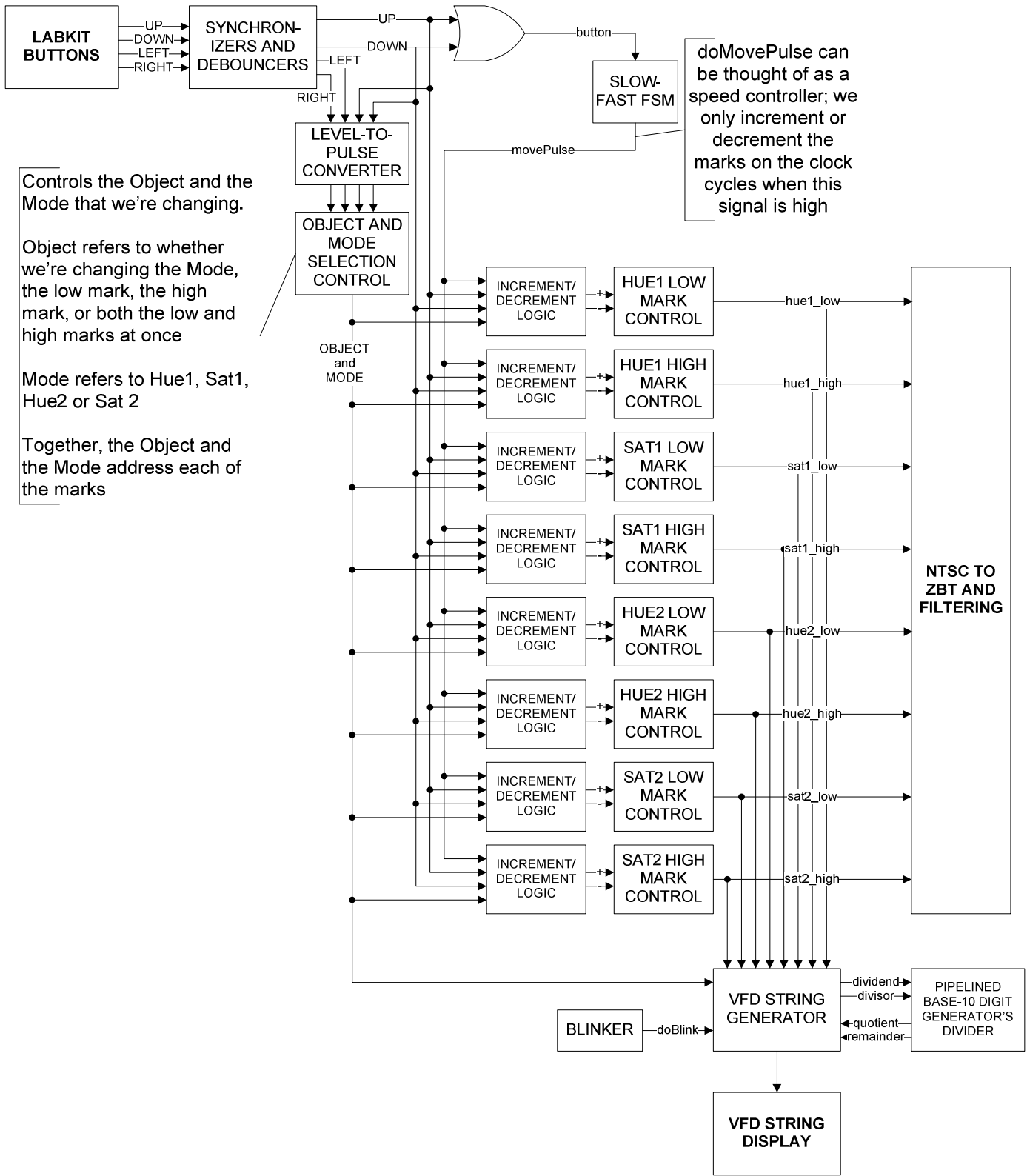
As described in the Filtering Overview section, each of the system's two color filters must have an adjustable Hue-Pass Low Mark, Hue-Pass High Mark, Saturation-Pass Low Mark, and Saturation Pass High Mark. These values are modified by the user on the fly using the Filter Configuration Module. The Filter Configuration Module uses the four directional push-buttons on the labkit as inputs and the labkit's VFD display as a feedback to the user.

Selection of the particular parameter of interest is accomplished using a menu-system on the VFD display. Each menu item on the VFD display is one of the two color filter's hue or saturation low and high marks. The VFD display can show one line of 16 characters, and the possible menu items are listed in Table 1:

Table 1: VFD Menu Options for Filter Configuration Parameter Selection

VFD Character #	0	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15
Item 1	h	1			l	o	w	x	x	x		h	i	x	x	x
Item 2	s	1			l	o	w	x	x	x		h	i	x	x	x
Item 3	h	2			l	o	w	x	x	x		h	i	x	x	x
Item 4	s	2			l	o	w	x	x	x		h	i	x	x	x

Each Menu Item will from here on be referred to as the Configuration Mode. Each item (Mode, Low-Mark, High-Mark) will be from here on referred to as the Configuration Object. Objects are selected by using the left and right buttons to move left or right. As a new Object is selected, its display changes from lower-case to upper-case and it begins blinking. The Object is changed by selecting it and pressing the up or down buttons. Because the range of the selection values is significant but precision is required in adjust the value, it is desirable to be able to go through the possible values both quickly and slowly. This is accomplished using a Slow/Fast FSM. When the FSM detects the user has pressed and holds a button, it initially increments or decrements the value by one at a slow interval. If the user keeps holding the button, it begins to increment or decrement the value rather quickly. Figure 6 shows a block diagram of the individual modules in the Filter Configuration Module, and Figure 7 is a state-transition diagram of the Slow/Fast FSM.



Controls the Object and the Mode that we're changing.

Object refers to whether we're changing the Mode, the low mark, the high mark, or both the low and high marks at once

Mode refers to Hue1, Sat1, Hue2 or Sat 2

Together, the Object and the Mode address each of the marks

doMovePulse can be thought of as a speed controller; we only increment or decrement the marks on the clock cycles when this signal is high

Figure 6: Block Diagram of Filter Configuration Module

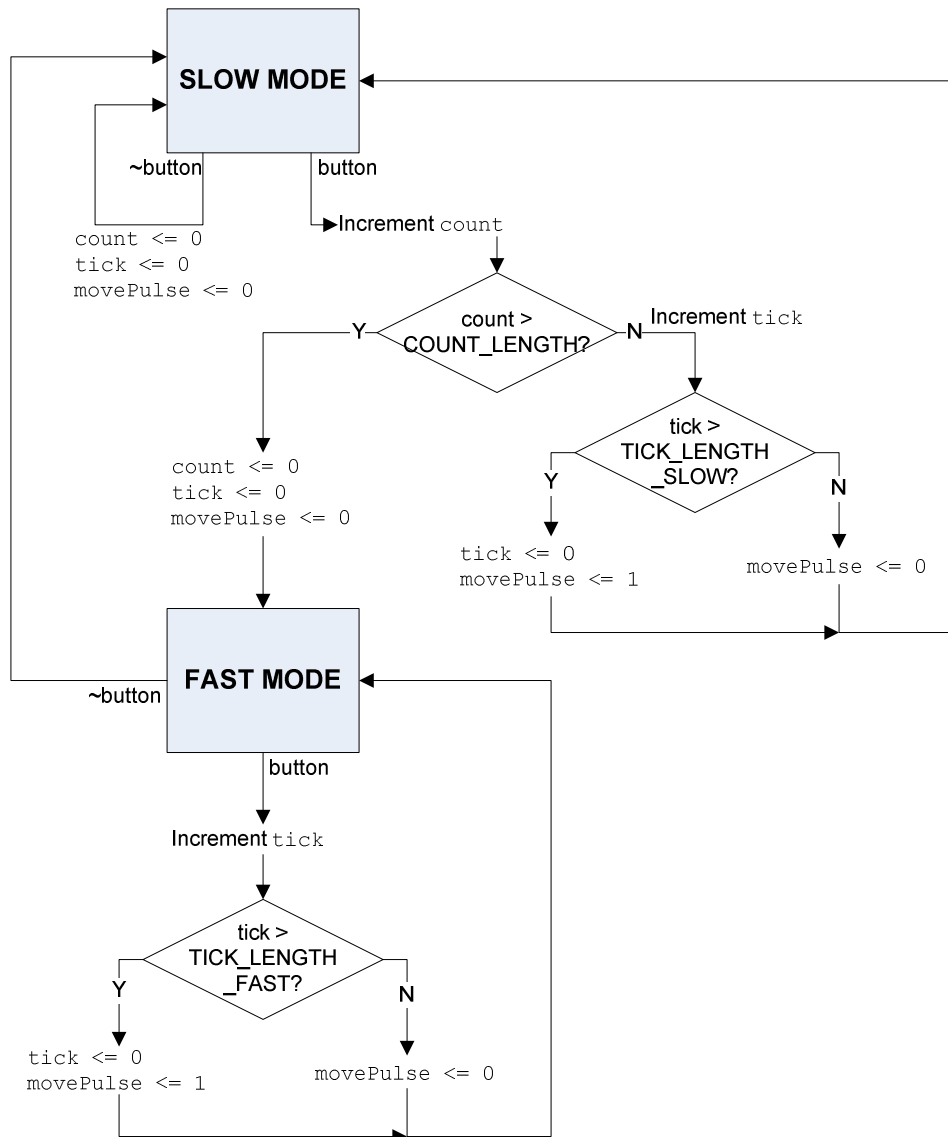


Figure 7: Slow-Fast FSM State Transition Diagram

Note: Transitions occur on the rising edge of each clock cycle. Shaded blocks are states. Non-shaded blocks are combinational logic pieces that determine the particular transition path.

The value of each mark is stored in the appropriate Mark Control Module. Each Mark Control Module has increment and decrement inputs. For every clock cycle in which its increment and decrement inputs are high, it increments or decrements its stored value appropriately. The Mark Control Verilog modules can be parameterized to use any minimum or maximum limits for their values, and they can be parameterized to use any value as their default start values. Thus, even though the Hue Marks go from 0 to 360 and the Saturation Marks go from 0 to 100, the same Verilog module was used but instantiated with different parameters.

Because the end-user experience was a priority for the filter customization, it was important that the VFD display show the filter parameters in common base-10 rather than a radix that only computer programmers would be familiar with. To generate the appropriate digits to be displayed, a divider module was used with a pipelined dividend/divisor multiplexer scheme. Figure 8 briefly illustrates the scheme used.

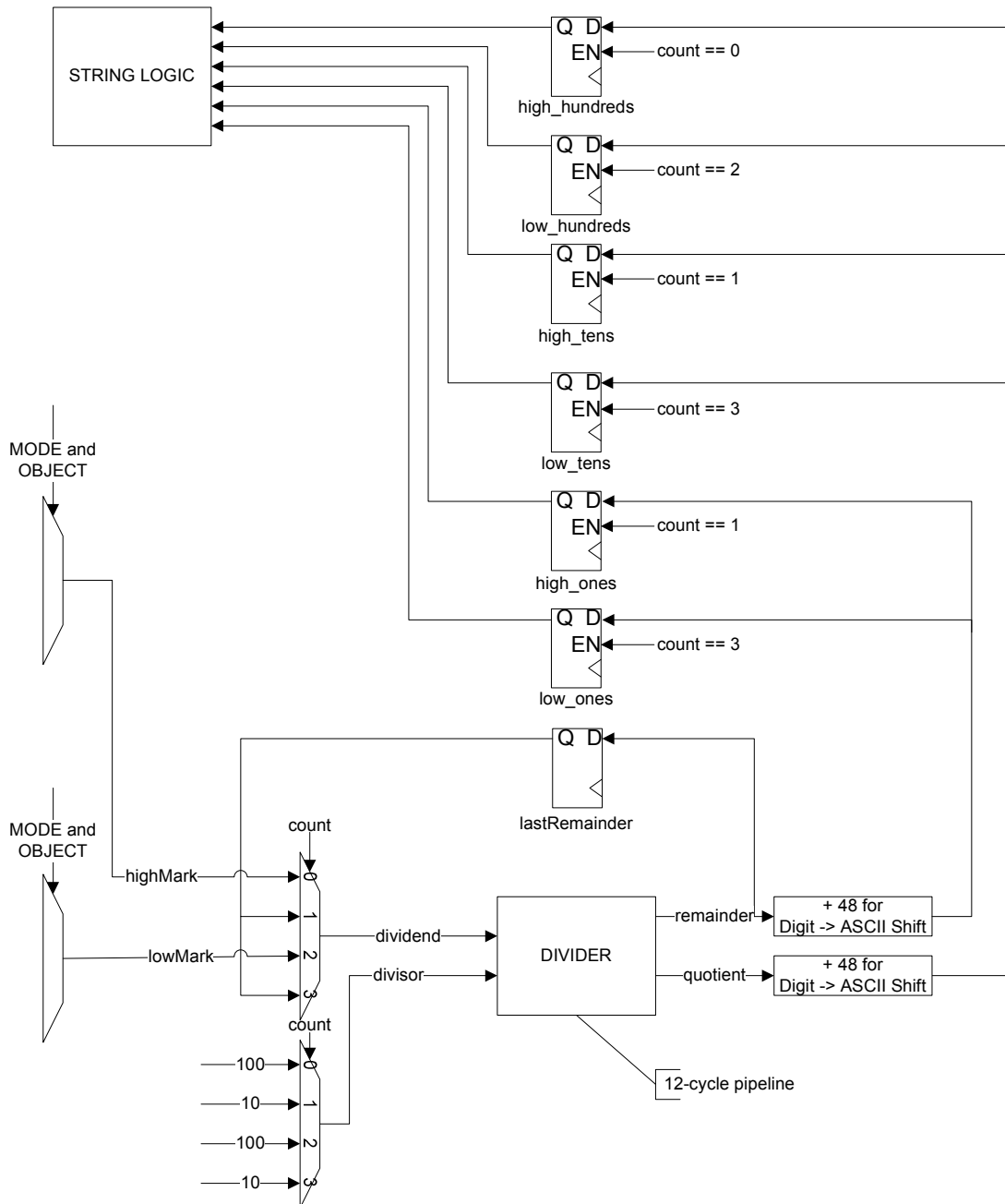


Figure 8: Pipelined Base-10 Digit Generator

NOTE: count increments on every clock cycle. highMark and lowMark are the current low and high marks of interest.

RGB To HSV Converter

Authorship: Daniel Lopuch

An RGB to HSV transformation is done using the following formulas:

$$H = \begin{cases} \text{undefined,} & \text{if } MAX = MIN \\ 60 \times \frac{G-B}{MAX-MIN} + 0, & \text{if } MAX = R \\ & \text{and } G \geq B \\ 60 \times \frac{G-B}{MAX-MIN} + 360, & \text{if } MAX = R \\ & \text{and } G < B \\ 60 \times \frac{B-R}{MAX-MIN} + 120, & \text{if } MAX = G \\ 60 \times \frac{R-G}{MAX-MIN} + 240, & \text{if } MAX = B \end{cases}$$
$$S = \begin{cases} 0, & \text{if } MAX = 0 \\ 1 - \frac{MIN}{MAX}, & \text{otherwise} \end{cases}$$

Image Source: Wikipedia (http://en.wikipedia.org/wiki/HSV_color_space)

Lacking a generalized ALU, each of the embedded operations were performed using an independent math module. In total, 4 adder/subtractors, 2 multipliers, and 2 dividers were used along with many registers that generated the appropriate synchronization delays to make a pipelined RGB to HSV Converter. Figure 9 shows a block-diagram of the implementation used. Note that because it was deemed unnecessary to add Value-checking to the filters, RGB to Value conversion is not implemented.

The total latency of the converter is 24 clock cycles. However, the converter is pipelined, so the throughput is still one conversion per clock cycle. The Filtering Module works by checking to see if the HSV values are within specified filter ranges, and if they are, it simply passes the RGB values through to the display pipeline. However, because the HSV results are delayed by 24 clock cycles, a 24-cycle pipelined delay was added to the RGB line to synchronize it with the results from the RGB to HSV transformation.

NTSC to ZBT and Filtering

Authorship: Daniel Lopuch

The NTSC to ZBT and Filtering module is based initially off of the “ZBT RAM Example” from the 2005 website. The example code does the conversion from YCrCb to RGB, generates the appropriate address, and synchronizes the RGB with the XVGA clock, and generates the appropriate write-enable signal when the new data is ready.

Some minor changes were done in the address generation to fit the new 1-pixel-per-word scheme. In addition, the appropriate delays were added to the RGB, address, and write-enable lines to synchronize them with the added HSV converter.

Filtering

In addition to the Hue-Pass and Saturation-Pass filters, a noise filter was added. During testing it was found that many small regions were picked up by the camera that happened to be inside the defined Color-Pass region. A noise filter was added that checks whether or not the last seven pixels passed the Color-Pass filters. If they did, and if the current pixel did as well, the current pixel passed the noise filter. This was implemented using a simple shift register for the Color-Pass result. When the color filters were tuned properly, the noise filter was effective at eliminating a lot of the noise without eliminating much of the regions of interest.

The filter-pass logic was wired such that each of the hue-pass, saturation-pass, and noise filters could be turned on or off using the labkit switches.

Centroid Calculation

The centroid, or center of mass, of each of the two filters was calculated by simply adding the x- and y-coordinates of each pixel that passed the respective filter to a separate sum. A counter kept track of how many pixels passed, and when the frame was over, and averager calculated the average. The delay in calculating the average was over 25 cycles at 60mhz, but because the value was updated only once per NTSC frame and because it was not really important *when* the average would update itself, this was not a problem.

The count of passing pixels was used later to determine which filter had the most passing pixels, and so which filter should be used to calculate the player's coordinates and determine whether the player is shooting or not.

Position Smoothing

During testing, it was observed that the centroids would jitter very quickly from frame to frame. A smoothing averager was added to address this problem. The smoother simply took the average of the current centroid position versus the average of the previous frame's centroid position. Mathematically, this results in an infinite average of past centroids, but each location further back in time is weighted less and less.

Output Debouncing

In testing, it was observed that the reported winning filter would sometimes quickly switch when the player's gesture did not change. The cause of this was determined to be slight increases of noise for one filter would temporarily push its filter's passing count above the other filter's passing count. A debouncer that sampled once per NTSC frame was added to debounce the winning filter output and eliminate false-positives.

Game Module

Authorship: Zachary Remscrim

The purpose of this module is to store the current state of the game, update the state of the game based on the current state and current inputs, and provide data to the display controller. The game state includes: the duck's position and heading, player's score, AI's score, number of shots the player has remaining this round, etc. The game state is updated once per frame of video (there are 60 frames of video per second). This is necessary since, if the state of the game were to change in the middle of a frame, part of the displayed frame would correspond to the old state, and the other part would correspond to the new state. This could, possibly, create a completely incorrect displayed image. For example, the duck could appear in two places at one time. To avoid this, state updates occur once per frame of video and are synchronized to the transition between frames of video. This is done by observing the vertical sync signal from the X VGA module, and executing a state transition whenever the vertical sync signal first goes low.

During a game state update, the states of all game objects are updated in parallel. The duck is moved to a new position, which is determined by its current position plus a small offset in the direction of its current heading. If this would move the duck outside of its valid region (which is bounded by the top, left, and right sides of the screen, as well as by the horizontal line which represents the upper boundary of the dirt image), then the duck would instead be reflected off the boundary. Since the duck is an animated sprite, the game module must also update the state of the animation. When the duck is flying, the animation has three possible states, which correspond to three different positions of the duck's wings. A given state is maintained for several frames of video, then the animation enters the next state. There are also animation states which correspond to the image to be displayed when the duck is first shot, as well as when the duck falls from the sky. These animation states are entered after the duck has been shot by either the player or the AI. In order to generate proper animation state information, a simple FSM was implemented in which each animation position was a state. State transitions are governed by the rules outlined above. This animation state information, as well as the duck's current position, is sent to the Duck Module (one of the graphics module) which will then return the proper image data. More details on this process will be included later in this section as well as in the section concerning the Duck Module.

Additionally, during a game state update, any shots fired by either the human player or the AI player are processed. If the firing player has no shots remaining this round, then nothing will occur. Otherwise, the number of shots that that player has will be reduced by one; and, if that player's cursor currently overlaps any portion of the duck, then the duck will be killed and that player's score will be updated accordingly. The score will be increased by a number of points inversely proportional to the time that has elapsed since the beginning of the round. For example, if less than 1 second has passed, the player will earn 300 points, while if 3 seconds have passed the player will only earn 100 points. Time is kept through instances of the RoundTimer Module. The score information and number of shots left will be sent to the ScoreOverlay module, which will be responsible for producing the appropriate display data.

A round can end in one of two ways. Either the duck can be killed, or the round timer could expire. If the duck has been killed, it will fall until it reaches the ground. If sufficiently long period of time has passed (specifically, 10 seconds) and neither the human player nor the AI player has successfully shot the duck, then the round will end and the duck will be removed from the screen. In either case, a special graphic will popup at the end of the round. If the player killed the duck, the popup will depict Cheney holding a dead duck. If the AI killed the duck, or the round ended due to time expiring, the popup will depict Cheney holding a gun. This will be accomplished by providing the RoundOverOverlay module with a signal that specifies which player killed the duck, if any, which will cause the RoundOverOverlay module to produce the appropriate display data. After a period of time has passed, the duck will return to the starting position, and a new round will begin. A game consists of a total of ten rounds, at which point the player can start a fresh game if they choose to.

As mentioned earlier, part of the purpose of this module is to provide data to the display controller. The Game Module does not produce this data on its own. Instead, it makes use of a series of graphics modules, each of which corresponds to a single game object, to produce the appropriate display information. These modules include the Duck Module, ScoreOverlay Module, and RoundOverOverlay Module mentioned already, as well as the TreeGrass Module, Background Module, and Crosshair Module. The game engine sends appropriate state information, as well as the coordinates of a pixel to these modules, and they each return the appropriate RGB color information, if the object they represent has a pixel at that location, to the Game Module. The Game Module then outputs the pixel that corresponds to the object closest to the foreground at the coordinates in question. Specific details concerning the operation of these modules can be found in the section dealing with the graphics pipeline.

AI Module

Authorship: Zachary Remscrim

The purpose of this module is to control the actions of the AI player. Specifically, this module will determine where the AI's crosshair will be moved at the next game state update, as well as whether or not the AI will fire at any given time. These decisions are sent to the game engine and are processed in an identical manner to the player's decision on movement and firing. This assures that the AI is given no special advantages or disadvantages. The game AI has the ability to be set to one of several difficulty settings, which affects how well the AI tracks the ducks movements and how accurate the AI's shots are.

To determine where to move the crosshair, the module examines the duck's current location and heading. It then estimates the location that the duck will be in at the next game state update. A small amount of error is purposely introduced here to cause the AI to not track the duck perfectly. This is done to make the AI's play more human-like. The amount of error varies by difficulty level. After the estimated location has been found, the AI will move its crosshair in the direction of the estimated location. The distance moved is limited to avoid the crosshair moving immediately to the duck's location. This limit also varies by difficulty level.

If the AI has any shots left (it has the same limit of three shots per round that the player has), then it will decide whether or not to shoot. If the location to which the AI has decided to move its crosshair at the next game state update places the crosshair over the duck's estimated position, and a sufficient amount of time has passed since the beginning of the round, then the AI will fire. Since the duck's estimated position had a small amount of error purposely introduced, it is possible that the AI will miss. Again, this is done to make the AI's play more human-like. The AI waits a small amount of time after a new round begins before it takes a shot in order to give the player more of a chance to kill the duck. The amount of time varies by difficulty level and is kept through an instance of the RoundTimer module.

Graphics Pipeline

Authorship: Zachary Remscrim

Due to the complexity of display data and the short clock period (15.4 nS), the various graphics modules mentioned earlier were organized into a graphics pipeline. They are each given the appropriate state information and values for hCount and vCount (the horizontal and vertical coordinates, respectively, of the pixel whose value is currently being requested) simultaneously, and compute the appropriate RGB value for that pixel in parallel (if the given module corresponds to an object that has a pixel at those coordinates). For a given hCount and vCount, the pixel color that will actually be displayed is the RGB value produced by the module highest in the pipeline that has color information for that pixel. This produces a layering effect whereby objects higher in the graphics pipeline are displayed above objects lower in the graphics pipeline. For example, the TreeGrass Module is above the Duck Module, which is above the Background Module in the graphics pipeline. Therefore, the tree and grass will be displayed over the duck, which in turn will be displayed

over the background. This will have the desired effect of having the duck appear as though it is flying behind the tree and grass, while still appearing above the sky.

Each module (with the exception of the background) consists of one or more single-port ROM modules, created using the Xilinx tools, that stores the bitmaps of the various objects. In general, these ROMs are 8 bits wide, and each entry stores the RGB values of a single pixel. For dichromatic images, such as the crosshair, the ROM is only 1 bit wide. In that case, each entry would be a 1 to represent one color, and a 0 to represent the other. The module stores the full 8 bit RGB values for both the “1” color and the “0” color, and returns the appropriate RGB value when either a 1 or 0 is encountered in the ROM. This was done to save memory. Each address corresponds to a single pixel of the bitmap. All RGB values are encoded in the 8 bit truecolor format, in which the high order 3 bits correspond to red, the next 3 bits correspond to green, and the low order 2 bits correspond to blue. Since the ADV7125 expects 24 bit color (8 bits for each of red, green and blue), an appropriate number of zeroes is added as low order bits to turn the 2 or 3 bit values into 8 bit values. For example, if a given pixel had the 8 bit RGB value 10111011, then 10100000, 11000000, and 11000000 would be the red, green, and blue values sent to the ADV7125. In many modules, the bitmaps stored in the ROMS were compressed relative to their display size. This was done to save memory. In order to display the image at the appropriate size, the same address in memory is accessed for several different hCount, vCount pairs.

In order to facilitate the creation of the .coe files needed by the ROMS, a Matlab script was written to convert .jpg, .gif, and .bmp files to .coe format. This made the graphics creation far easier since it eliminated the otherwise absurdly tedious task of generating .coe files.

The modules that make up the graphics pipeline are: Duck, Crosshair, TreeGrass, Background, ScoreOverlay, RoundOverOverlay, and TitleScreen.

Duck Module

This module is responsible for displaying the animated duck. It possesses 6 ROMs, each of which corresponds to one of the 6 possible duck images. 3 of these images form the flying animation, 2 form the falling animation, and the last is displayed when the duck is first shot. The Game Module sends a state signal to this module to specify which image should be displayed. For example, when the duck is in its standard flying state, the state signal repeatedly cycles through the 3 flying images. Thus, it appears as though the duck is flapping its wings while flying. The Game Module also provides the current coordinates of the top left corner of the duck. This allows the Duck Module know where the duck is, and provide RGB information for a given hCount and vCount.

Crosshair Module

This module is responsible for displaying a crosshair. There are two instances of this module in the device, one for the player’s crosshair, and another for the AI’s

crosshair. Since the bitmap for a crosshair is dichromatic (the “X” portion of the crosshair is one color, and the background is another color), the ROM that stores the crosshair bitmap is only 1 bit wide, for the reason discussed above. This module has a color parameter, which specifies the color to be returned whenever a portion of the “X” is encountered in memory. This parameter is set to a different value in each instance of the module, thereby allowing the two crosshairs to be distinguished. No color information is returned when the background is encountered. This is done to allow the image of whatever the crosshair is over to still be visible, while still accurately displaying the “X” portion. As was the case for the Duck Module, this module needs the coordinates of the object that it represents. In this case, the Game Module transmits the x and y coordinates of the center of the “X”. Since the crosshair is not animated, no state information is needed.

Tree and Grass Modules

This module stores the color information for the tree, grass, and dirt. Since all these objects are static, neither state information nor coordinates are required. For convenience, the bitmap of the tree is stored in a different ROM than the bitmap of the grass and dirt.

Background Module

Unlike the other modules that make up the graphics pipeline, this module has no need for a ROM. This module simply returns a blue pixel for any hCount, vCount pair in the sky, and a brown pixel otherwise. If the device is working properly, this brown pixel will never be displayed since the grass and dirt are displayed over it. It was included to assure that something reasonable is displayed even if there is a glitch in the TreeGrass Module.

Score Overlay Module

This module is responsible for displaying all score related information, as well as the number of shots that the player has left. The score information includes the player’s score and AI’s score, which are computed by the Game Module. Additionally, there is a display of the 10 ducks that appear in a round. The color of the duck corresponds to its state. White indicates that the duck has not been killed, red indicates that it was killed by the AI, and green indicates that it was killed by the player. This data is provided by the Game Module.

Round-Over Overlay Module

This module is responsible for displaying a graphic popup whenever the round ends. The graphic popup is either a fake image of Dick Cheney brandishing a rifle or a fake image of Dick Cheney holding a dead duck. The Game Module sends signals to this module indicating the player that killed the duck, or that the duck was not killed, as well location of the top left corner of the Dick Cheney image (which is needed since the image slowly rises from behind the grass). If the human player killed the duck, the popup will show Cheney holding a dead duck. If the AI player shot the duck, or no player shot the duck, the popup will show Cheney holding a rifle.

Title Screen Module

Authorship: Zachary Remscrim

This module displays the title screen. It receives a signal from the Game Module which specifies whether or not the title screen should be displayed at any point in time.

Round Timer Module

Authorship: Zachary Remscrim

The purpose of this module is to keep track of the passage of time. Upon receiving a start signal, the module will increment a counter variable at each clock edge until that variable reaches a threshold, at which point it will assert a time expired signal. The time expired signal will remain asserted until this module is reset. This threshold is determined by multiplying the clock frequency by the length of the desired timing interval. The length of the timing interval is a parameter that can be set on instantiation. This allows different instances of the module to measure different lengths of time. Several instances of this module are present throughout the device. For example, the Game Module uses an instance of this module to determine when the time limit of a round has expired.

TESTING AND DEBUGGING

Video Decoding and Processing

When the project was first conceived, the video filtering was simply a single hue-pass region. Testing, however, showed this was a dramatic oversimplification. The first problem that became apparent during testing was how insufficient a hue-pass filter region was by itself. As described in the Filtering Overview section, it was quickly learned that regions of near-white pixels tended to display as pixels of all hues; when the camera was focused on a white area, there would always be noise from there regardless of which hue was being passed through the filter. A second filter – the Saturation-Pass Filter – improved this problem significantly.

It was also realized early on that the customization of the filter parameters was a necessity rather than a convenience. Depending on the particular lab bench chosen for the day and the position of the sun through the outside windows, the lighting conditions would change enough that if the camera was calibrated for a certain color at one bench, at another bench or at a different time of day the perceived color would change enough to make the calibration outdated.

A further complication in calibrating to perceived colors came from the camera itself. The camera that the group was assigned had built-in gain and color-shift circuitry. These are desirable features for a consumer electronics device since they ensure the best picture based on changing lighting conditions, but for us, this

unnecessarily complicated the project. We would find that we could calibrate the camera's detection to work well for one region, but when we would move our hand closer to it or to another position (such as blocking a ceiling light from the camera's view), the camera's auto-gain would kick in and shift perceived colors out of our calibration region. We experimented with other groups' cameras, and we found that other groups cameras lacked this auto-gain function and so performed significantly better.

Another grievance that was had with the assigned camera was the lack of picture quality. One issue that was noticed right away was for the camera to shift the picture towards green. Because it lacked significant color depth, any gray or dark pixels would often be shifted into the green hue region, and so if the filter was calibrated for green, there would be a significant amount of noise. Again, other groups' cameras, although older, proved to have a much greater color depth and noise tended to be inside the color's actual hue region rather than shifted into the green region.

Both types of cameras, however, displayed significant trouble with sharp edges. When they would see a sharp edge, they would actually register pixels colored with colors that we not actually in the edge but just showed up as noise. Dealing with the noise encouraged the development of the noise filter, but an ultimately more effective and low-tech solution was found in simply blurring the camera's focus – all we cared about was regions of color, and blurring the camera's focus effectively eliminated edge-noise.

Game Engine Debugging

In order to facilitate easy debugging of the Game Module, AI Module, and graphics modules, careful attention was paid to assure their modularity. This greatly simplified testing since each module could be tested independently. During the early stages of the project, work was focused on the creation of the Game Module and the framework of the graphics modules. Initially, many of the bitmaps used by the graphic modules were simply single color rectangles that were used purely for testing purposes to verify that the module correctly displayed a simple image at the times dictated by the Game Module. Additionally, the input that would normally come from the Image Processing Module was replaced by a simple input through the labkit's buttons This allowed the basic functionality of the Game Module and graphics modules to be tested. Once this stage of the project was fully debugged, actual bitmaps were used by the graphics modules, the AI was introduced, and the Game Module had all desired functionality implemented. After the debugging this stage, the final phase of the project began. This included integration of these modules with the Image Processing Module, as well as improving the general look and performance of the game.

While the initial Game Module and basic graphic modules were mostly bug free, many bugs were encountered once the AI was introduced. The initial behavior of the AI was problematic. It had a tendency to wander, not accurately track the duck, and

fire at strange times. After a closer examination of the AI module, including extensive use of the logic analyzer to examine the critical signals of this module, it was discovered that the AI's calculation for the duck's position had more error than intended (a small amount of error is purposely present, in order to make the AI's play more realistic). Once this error was discovered, it was quickly corrected, and many of the bugs present disappeared.

CONCLUSIONS

The goal of this project was to create a device that enabled a user to play a game of Duck Hunt, with several added features. These features included the addition of a "Realistic Dick Cheney AI" which competes against the player as well as the replacement of the traditional "Duck Hunt Gun" with an input system based on moving one's hand in front of a camera.

After extensive testing of all modules, it is clear that these goals have been satisfied. With the exception of a few minor, and rare, glitches, as well as a slight jitter presence in the input system, all aspects of the device behave as desired. All game rules are accurately followed, all graphics are displayed properly, the AI functions well, and visual inputs are successfully delivered to the game.

The design and implementation of this device demonstrated the value of modularity in digital systems (or any other system, for that matter). By separating the functionality of the device into several modules, each module could be tested and debugged independently. This greatly sped up the debugging process.

Furthermore, the modular nature of the device makes the system easier to understand due to the layers of abstraction it provides. The system as a whole can be analyzed from a global perspective without considering the specific implementation details of a given module. This has the further advantage of allowing a module to be revised for the purpose of improving efficiency. Any such internal change will not affect the overall behavior of the device if the utilized abstractions are used.

APPENDIX 1: VERILOG CODE

Base Labkit File

```
////////////////////////////////////
//
// 6.111 FPGA Labkit -- Template Toplevel Module
//
// For Labkit Revision 004
//
//
// Created: October 31, 2004, from revision 003 file
// Author: Nathan Ickes
//
////////////////////////////////////
//
// CHANGES FOR BOARD REVISION 004
//
// 1) Added signals for logic analyzer pods 2-4.
// 2) Expanded "tv_in_ycrcb" to 20 bits.
// 3) Renamed "tv_out_data" to "tv_out_i2c_data" and "tv_out_sclk" to
//     "tv_out_i2c_clock".
// 4) Reversed disp_data_in and disp_data_out signals, so that "out" is an
//     output of the FPGA, and "in" is an input.
//
// CHANGES FOR BOARD REVISION 003
//
// 1) Combined flash chip enables into a single signal, flash_ce_b.
//
// CHANGES FOR BOARD REVISION 002
//
// 1) Added SRAM clock feedback path input and output
// 2) Renamed "mousedata" to "mouse_data"
// 3) Renamed some ZBT memory signals. Parity bits are now incorporated into
//     the data bus, and the byte write enables have been combined into the
//     4-bit ram#_bwe_b bus.
// 4) Removed the "systemace_clock" net, since the SystemACE clock is now
//     hardwired on the PCB to the oscillator.
//
////////////////////////////////////
//
// Complete change history (including bug fixes)
//
// 2005-Sep-09: Added missing default assignments to "ac97_sdata_out",
//              "disp_data_out", "analyzer[2-3]_clock" and
//              "analyzer[2-3]_data".
//
// 2005-Jan-23: Reduced flash address bus to 24 bits, to match 128Mb devices
//              actually populated on the boards. (The boards support up to
//              256Mb devices, with 25 address lines.)
//
// 2004-Oct-31: Adapted to new revision 004 board.
//
// 2004-May-01: Changed "disp_data_in" to be an output, and gave it a default
//              value. (Previous versions of this file declared this port to
//              be an input.)
//
// 2004-Apr-29: Reduced SRAM address busses to 19 bits, to match 18Mb devices
//              actually populated on the boards. (The boards support up to
//              72Mb devices, with 21 address lines.)
//
//
```

```

// 2004-Apr-29: Change history started
//
////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////
module DuckHunt    (beep, audio_reset_b, ac97_sdata_out, ac97_sdata_in,
ac97_synch,
                    ac97_bit_clock,

                    vga_out_red, vga_out_green, vga_out_blue, vga_out_sync_b,
                    vga_out_blank_b, vga_out_pixel_clock, vga_out_hsync,
                    vga_out_vsync,

                    tv_out_ycrCb, tv_out_reset_b, tv_out_clock, tv_out_i2c_clock,
                    tv_out_i2c_data, tv_out_pal_ntsc, tv_out_hsync_b,
                    tv_out_vsync_b, tv_out_blank_b, tv_out_subcar_reset,

                    tv_in_ycrCb, tv_in_data_valid, tv_in_line_clock1,
                    tv_in_line_clock2, tv_in_aef, tv_in_hff, tv_in_aff,
                    tv_in_i2c_clock, tv_in_i2c_data, tv_in_fifo_read,
                    tv_in_fifo_clock, tv_in_iso, tv_in_reset_b, tv_in_clock,

                    ram0_data, ram0_address, ram0_adv_ld, ram0_clk, ram0_cen_b,
                    ram0_ce_b, ram0_oe_b, ram0_we_b, ram0_bwe_b,

                    ram1_data, ram1_address, ram1_adv_ld, ram1_clk, ram1_cen_b,
                    ram1_ce_b, ram1_oe_b, ram1_we_b, ram1_bwe_b,

                    clock_feedback_out, clock_feedback_in,

                    flash_data, flash_address, flash_ce_b, flash_oe_b, flash_we_b,
                    flash_reset_b, flash_sts, flash_byte_b,

                    rs232_txd, rs232_rxd, rs232_rts, rs232_cts,

                    mouse_clock, mouse_data, keyboard_clock, keyboard_data,

                    clock_27mhz, clock1, clock2,

                    disp_blank, disp_data_out, disp_clock, disp_rs, disp_ce_b,
                    disp_reset_b, disp_data_in,

                    button0, button1, button2, button3, button_enter, button_right,
                    button_left, button_down, button_up,

                    switch,

                    led,

                    user1, user2, user3, user4,

                    daughtercard,

                    systemace_data, systemace_address, systemace_ce_b,
                    systemace_we_b, systemace_oe_b, systemace_irq, systemace_mpbdrdy,

                    analyzer1_data, analyzer1_clock,
                    analyzer2_data, analyzer2_clock,
                    analyzer3_data, analyzer3_clock,
                    analyzer4_data, analyzer4_clock);

output beep, audio_reset_b, ac97_synch, ac97_sdata_out;
input  ac97_bit_clock, ac97_sdata_in;

```

```

output [7:0] vga_out_red, vga_out_green, vga_out_blue;
output vga_out_sync_b, vga_out_blank_b, vga_out_pixel_clock,
      vga_out_hsync, vga_out_vsync;

output [9:0] tv_out_ycrCb;
output tv_out_reset_b, tv_out_clock, tv_out_i2c_clock, tv_out_i2c_data,
      tv_out_pal_ntsc, tv_out_hsync_b, tv_out_vsync_b, tv_out_blank_b,
      tv_out_subcar_reset;

input  [19:0] tv_in_ycrCb;
input  tv_in_data_valid, tv_in_line_clock1, tv_in_line_clock2, tv_in_aef,
      tv_in_hff, tv_in_aff;
output tv_in_i2c_clock, tv_in_fifo_read, tv_in_fifo_clock, tv_in_iso,
      tv_in_reset_b, tv_in_clock;
inout  tv_in_i2c_data;

inout  [35:0] ram0_data;
output [18:0] ram0_address;
output ram0_adv_ld, ram0_clk, ram0_cen_b, ram0_ce_b, ram0_oe_b, ram0_we_b;
output [3:0] ram0_bwe_b;

inout  [35:0] ram1_data;
output [18:0] ram1_address;
output ram1_adv_ld, ram1_clk, ram1_cen_b, ram1_ce_b, ram1_oe_b, ram1_we_b;
output [3:0] ram1_bwe_b;

input  clock_feedback_in;
output clock_feedback_out;

inout  [15:0] flash_data;
output [23:0] flash_address;
output flash_ce_b, flash_oe_b, flash_we_b, flash_reset_b, flash_byte_b;
input  flash_sts;

output rs232_txd, rs232_rts;
input  rs232_rxd, rs232_cts;

input  mouse_clock, mouse_data, keyboard_clock, keyboard_data;

input  clock_27mhz, clock1, clock2;

output disp_blank, disp_clock, disp_rs, disp_ce_b, disp_reset_b;
input  disp_data_in;
output disp_data_out;

input  button0, button1, button2, button3, button_enter, button_right,
      button_left, button_down, button_up;
input  [7:0] switch;
output [7:0] led;

inout [31:0] user1, user2, user3, user4;

inout [43:0] daughtercard;

inout  [15:0] systemace_data;
output [6:0] systemace_address;
output systemace_ce_b, systemace_we_b, systemace_oe_b;
input  systemace_irq, systemace_mpbardy;

output [15:0] analyzer1_data, analyzer2_data, analyzer3_data,
      analyzer4_data;
output analyzer1_clock, analyzer2_clock, analyzer3_clock, analyzer4_clock;

```



```

////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////
//
// I/O Assignments
//
////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////

// Audio Input and Output
assign beep= 1'b0;
assign audio_reset_b = 1'b0;
assign ac97_synch = 1'b0;
assign ac97_sdata_out = 1'b0;
/*
*/
// ac97_sdata_in is an input

// Video Output
assign tv_out_ycrcb = 10'h0;
assign tv_out_reset_b = 1'b0;
assign tv_out_clock = 1'b0;
assign tv_out_i2c_clock = 1'b0;
assign tv_out_i2c_data = 1'b0;
assign tv_out_pal_ntsc = 1'b0;
assign tv_out_hsync_b = 1'b1;
assign tv_out_vsync_b = 1'b1;
assign tv_out_blank_b = 1'b1;
assign tv_out_subcar_reset = 1'b0;

// Video Input
//assign tv_in_i2c_clock = 1'b0;
assign tv_in_fifo_read = 1'b1;
assign tv_in_fifo_clock = 1'b0;
assign tv_in_iso = 1'b1;
//assign tv_in_reset_b = 1'b0;
assign tv_in_clock = clock_27mhz;//1'b0;
//assign tv_in_i2c_data = 1'bZ;
// tv_in_ycrcb, tv_in_data_valid, tv_in_line_clock1, tv_in_line_clock2,
// tv_in_aef, tv_in_hff, and tv_in_aff are inputs

// SRAMs

/* change lines below to enable ZBT RAM bank0 */

/*
assign ram0_data = 36'hZ;
assign ram0_address = 19'h0;
assign ram0_clk = 1'b0;
assign ram0_we_b = 1'b1;
assign ram0_cen_b = 1'b0;      // clock enable
*/

/* enable RAM pins */

assign ram0_ce_b = 1'b0;
assign ram0_oe_b = 1'b0;
assign ram0_adv_ld = 1'b0;
assign ram0_bwe_b = 4'h0;

/*****/

assign ram1_data = 36'hZ;
assign ram1_address = 19'h0;
assign ram1_adv_ld = 1'b0;
assign ram1_clk = 1'b0;

```

```

assign raml_cen_b = 1'b1;
assign raml_ce_b = 1'b1;
assign raml_oe_b = 1'b1;
assign raml_we_b = 1'b1;
assign raml_bwe_b = 4'hF;

assign clock_feedback_out = 1'b0;
// clock_feedback_in is an input

// Flash ROM
assign flash_data = 16'hZ;
assign flash_address = 24'h0;
assign flash_ce_b = 1'b1;
assign flash_oe_b = 1'b1;
assign flash_we_b = 1'b1;
assign flash_reset_b = 1'b0;
assign flash_byte_b = 1'b1;
// flash_sts is an input

// RS-232 Interface
assign rs232_txd = 1'b1;
assign rs232_rts = 1'b1;
// rs232_rxd and rs232_cts are inputs

// PS/2 Ports
// mouse_clock, mouse_data, keyboard_clock, and keyboard_data are inputs

// LED Displays
/*
assign disp_blank = 1'b1;
assign disp_clock = 1'b0;
assign disp_rs = 1'b0;
assign disp_ce_b = 1'b1;
assign disp_reset_b = 1'b0;
assign disp_data_out = 1'b0;
*/
// disp_data_in is an input

// Buttons, Switches, and Individual LEDs
//lab3 assign led = 8'hFF;
// button0, button1, button2, button3, button_enter, button_right,
// button_left, button_down, button_up, and switches are inputs

// User I/Os
assign user1 = 32'hZ;
assign user2 = 32'hZ;
assign user3 = 32'hZ;
assign user4 = 32'hZ;

// Daughtercard Connectors
assign daughtercard = 44'hZ;

// SystemACE Microprocessor Port
assign systemace_data = 16'hZ;
assign systemace_address = 7'h0;
assign systemace_ce_b = 1'b1;
assign systemace_we_b = 1'b1;
assign systemace_oe_b = 1'b1;
// systemace_irq and systemace_mpbrdy are inputs

// Logic Analyzer
/*assign analyzer1_data = 16'h0;
assign analyzer1_clock = 1'b1;

```

```

assign analyzer2_data = 16'h0;
assign analyzer2_clock = 1'b1;
assign analyzer3_data = 16'h0;
assign analyzer3_clock = 1'b1;
assign analyzer4_data = 16'h0;
assign analyzer4_clock = 1'b1;*/

// use FPGA's digital clock manager to produce a
// 65MHz clock (actually 64.8MHz)
wire clock_65mhz_unbuf,clock_65mhz;
DCM vclk1(.CLKIN(clock_27mhz),.CLKFX(clock_65mhz_unbuf));
// synthesis attribute CLKFX_DIVIDE of vclk1 is 10
// synthesis attribute CLKFX_MULTIPLY of vclk1 is 24
// synthesis attribute CLK_FEEDBACK of vclk1 is NONE
// synthesis attribute CLKIN_PERIOD of vclk1 is 37
BUFG vclk2(.O(clock_65mhz),.I(clock_65mhz_unbuf));
wire clk = clock_65mhz;
// power-on reset generation
wire power_on_reset; // remain high for first 16 clocks
SRL16 reset_sr (.D(1'b0), .CLK(clock_65mhz), .Q(power_on_reset),
.A0(1'b1), .A1(1'b1), .A2(1'b1), .A3(1'b1));
defparam reset_sr.INIT = 16'hFFFF;

// ENTER button is user reset
wire reset,user_reset;
debounce dbl(power_on_reset, clock_65mhz, ~button_enter, user_reset);
assign reset = user_reset | power_on_reset;

reg [7:0] rgb;

// generate basic XVGA video signals
wire [10:0] hcount, hcount_inverse;
assign hcount_inverse = hcount <= 720 ? 720 - hcount : 0;
wire [9:0] vcount;
wire hsync,vsync,blank;
xvga xvga1(clk,hcount,vcount,hsync,vsync,blank);

////////////////////
//HUE LIMITS CONTROLLER
// This part makes the hue-pass controlled by the labkit buttons

wire [8:0] hue1_lowMark, hue1_highMark, hue2_lowMark, hue2_highMark;
//hue values between 0 and 360 to pass
wire [6:0] sat1_lowMark, sat1_highMark, sat2_lowMark, sat2_highMark;
wire [16*8-1:0] string_data_out;
reg [16*8-1:0] string_data[1:0];
limits_interface li(clk, 1'b0, ~button_up, ~button_down, ~button_left,
~button_right, button_enter, //don't invert enter_button so that stuff happens
on buttonUp rather than on buttonDown
hue1_lowMark,
hue1_highMark, hue2_lowMark, hue2_highMark,
sat1_lowMark,
sat1_highMark, sat2_lowMark, sat2_highMark,
string_data_out);

```

```

        //synchronize data to 27mhz clock
        always @ (posedge clock_27mhz) begin
            {string_data[1], string_data[0]} <= {string_data[0],
string_data_out};
        end

        assign led = hue1_lowMark;

        // END HUE LIMITS
        //////////////////////////////////////

        display_string ds(reset, clock_27mhz, string_data[1],
disp_clock, disp_rs, disp_ce_b, disp_data_out);
                                                                    disp_blank,
                                                                    disp_reset_b,

        // wire up to ZBT ram

        wire [35:0] vram_write_data;
        wire [35:0] vram_read_data;
        wire [18:0] vram_addr;
        wire          vram_we;

        zbt_6111 zbt1(clk, 1'b1, vram_we, vram_addr,
                    vram_write_data, vram_read_data,
                    ram0_clk, ram0_we_b, ram0_address, ram0_data, ram0_cen_b);

        // generate pixel value from reading ZBT memory
        wire [23:0] vr_pixel; //Full RGB info
        wire [18:0] vram_addr1;

        vram_display vd1(reset, clk, hcount_inverse, vcount, vr_pixel,
                    vram_addr1, vram_read_data);

        // ADV7185 NTSC decoder interface code
        // adv7185 initialization module
        adv7185init adv7185(.reset(reset), .clock_27mhz(clock_27mhz),
                    .source(1'b0), .tv_in_reset_b(tv_in_reset_b),
                    .tv_in_i2c_clock(tv_in_i2c_clock),
                    .tv_in_i2c_data(tv_in_i2c_data));

        wire [29:0] ycrCb; // video data (luminance, chrominance)
        wire [2:0] fvh; // sync for field, vertical, horizontal
        wire          dv; // data valid

        ntsc_decode decode (.clk(tv_in_line_clock1), .reset(reset),
                    .tv_in_ycrCb(tv_in_ycrCb[19:10]),
                    .ycrCb(ycrCb), .f(fvh[2]),
                    .v(fvh[1]), .h(fvh[0]), .data_valid(dv));

        // code to write NTSC data to video memory

        wire [18:0] ntsc_addr;
        wire [35:0] ntsc_data;
        wire          ntsc_we;
        wire [9:0] x1_avg, y1_avg, x2_avg, y2_avg;
        wire filterWinner; //Which filter is winning, 0==filter1, 1==filter2

```

```

/*DEBUGS:
  wire [26:0] x_sum;
  wire [26:0] new_x_avg;
  wire [18:0] pixelCount;
  wire startFinalAvgDividing, endFinalAvgDividing;
  wire [9:0] x;
  wire [7:0] y; */

  wire [18:0] pixelCount1, pixelCount2;

  ntsc_to_zbt n2z (clk, tv_in_line_clock1, fvh, dv, ycrCb,
    ntsc_addr, ntsc_data, ntsc_we,
    1'b0, switch[5],
    hue1_lowMark, hue1_highMark, hue2_lowMark, hue2_highMark,
    sat1_lowMark, sat1_highMark, sat2_lowMark, sat2_highMark,
    switch[4], switch[3], switch[2], switch[1],
    x1_avg, x2_avg, y1_avg, y2_avg, filterWinner, 1,
switch[0],
    pixelCount1, pixelCount2 );
    //x_sum, new_x_avg, pixelCount, startFinalAvgDividing,
endFinalAvgDividing, x, y);

// code to write pattern to ZBT memory
reg [31:0] count;
always @(posedge clk) count <= reset ? 0 : count + 1;

wire [18:0] vram_addr2 = count[0+18:0];
wire [35:0] vpat = ( switch[1] ? {4{count[3+3:3]},4'b0}
  : {4{count[3+4:4]},4'b0} );

  wire sw_ntsc = 1'b1;
  wire [18:0] write_addr = sw_ntsc ? ntsc_addr : vram_addr2;
  wire [35:0] write_data = sw_ntsc ? ntsc_data : vpat;

  //SWITCH BETWEEN READING AND WRITING MODES:

  /*Method 1:
    Give vram priority to xvga
    Here we write new data only when the current xvga pixel is
outside of the NTSC area.
    Result is you get "waves" of old frames in the display
since we can't write new video data
    constantly. We take advantage of the fact that XVGA clock
and video clock are not multiples,
    so because of non-constant interference, we do write new
video data to all pixels eventually,
    just not in the same frame.
  wire read_select = (hcount <= 720) && (vcount <= 480);
  //when this wire is 1, we want to reader.

                                                                    //when this
wire is 0, we want to write.
  assign vram_addr = (read_select) ? vram_addr1 : write_addr;
  assign vram_write_data = write_data;

```

```

assign vram_we = ~read_select;

//NTSC Mask (during write mode, display b/w bars)
//wire [23:0] pixel = vr_pixel;
reg [23:0] pixel;
always @(posedge clk)
begin
    //pixel <= switch[5] ? vr_pixel : {hcount[8:6],5'b0,
hcount[8:6],5'b0, hcount[8:6],5'b0};
    pixel <= (hcount <= 723 && vcount <= 483) ? vr_pixel : 24'b0;
end

/**/

/*Method 2:
    Give vram priority to video
    Here we write new data whenever it arrives. The problem is
that certain xvga pixels will not be
    able to be reader. The advantage here is that we don't get
"waves of old data" as in method 1.
    The disadvantage is we get bits of random snow scattered
throughout (ie at those pixels where new
    data is being written).
    We'll hack around the snow by when data is being written,
we'll just display the previous pixel again.
    Thus the displayed image won't be true, but it shouldn't be
noticeable since we're repeating only
    single pixels. */
assign vram_addr = (ntsc_we) ? write_addr : vram_addr1;
assign vram_write_data = write_data;
assign vram_we = ntsc_we; /**/

//Delay the we by ZBT read + color conversion so we know when we hit a
"snow pixel"
wire we_delay;
//wire theDelayedWE;
//delayN we_delayer(clk, vram_we, theDelayedWE);
//defparam we_delayer.NDELAY = 2; //2 for ZBT pipeline
reg delayed_we[1:0];
always @(posedge clk)
    {delayed_we[1], delayed_we[0]} <= {delayed_we[0], vram_we};

wire theDelayedWE = delayed_we[1]; /**/

//NTSC Mask (during write mode, display b/w bars)
//wire [23:0] pixel = vr_pixel;
reg [23:0] pixel;
wire [23:0] pixel_out[2:0];
reg [23:0] oldPixel;
always @(posedge clk)
begin
    //pixel <= switch[5] ? vr_pixel : {hcount[8:6],5'b0,
hcount[8:6],5'b0, hcount[8:6],5'b0};
    oldPixel <= vr_pixel;
    pixel <= (hcount <= 722 && vcount <= 482) ?
        (theDelayedWE ? oldPixel : vr_pixel) :
//When the pixel was being written to (ie couldn't read it), interpolate by
putting in the previous pixel
        24'b0;
//else (when we're outside the NTSC box),
display blackness

```

```

end

//DRAWING FILTER TRACKERS
wire [8*3-1:0] filter1_tracker, filter2_tracker;
//filterWinner is which color filter has more pixels. ==0 -> filter1, ==1
-> filter2
assign filter1_tracker = (~filterWinner) ?
                                {8'd0,
8'd255, 8'b0} : {8'd255, 8'd255, 8'b0};
assign filter2_tracker = (filterWinner) ?
                                {8'd0,
8'd255, 8'b0} : {8'd255, 8'd255, 8'b0};

assign pixel_out[0] = (hcount_inverse == {1'b0, x1_avg} || vcount ==
y1_avg) ? filter1_tracker : pixel;
assign pixel_out[1] = (hcount_inverse == {1'b0, x2_avg} || vcount ==
y2_avg) ? filter2_tracker : pixel_out[0];

//GENERATING 1024x768-TRANSLATED X AND Y AVERAGES (only winning x & y)
wire [9:0] x720, y720, x1024, y1024;
assign x720 = filterWinner ? //filterWinner is which color filter has
more pixels. ==0 -> filter1, ==1 -> filter2
                                (720 - x2_avg) : (720 - x1_avg);
assign y720 = filterWinner ?
                                y2_avg : y1_avg;

// feed XVGA and AI signals to game
wire [7:0] aPixel;
wire pbsync,pvsync,pblank,AIFire;
wire [10:0] AICrosshairX, duckX, x720Prime;
wire [9:0] AICrosshairY, duckY, y720Prime;
wire duckAlive,roundOver,roundStart,inCalibrate;
wire [1:0] playerShotsLeft,AIShotsLeft,dir;
wire
abutton_left,abutton_right,abutton_up,abutton_down,aButton2,aButton3,calibrate,
startPlaying;

assign pixel_out[2] = (hcount == x720Prime || vcount == y720Prime) ?
{8'b0, 8'b0, 8'd255} : pixel_out[1];

wire hsPrime,vsPrime,bPrime;
delayN dn1(clk,hsync,hsPrime); // delay by 3+2+1 cycles to sync with ZBT read
+ YCrCb2RGB + RGB2Hue + NTSC mask
delayN dn2(clk,vsync,vsPrime);
delayN dn3(clk,blank,bPrime);
defparam dn1.NDELAY = 4;
defparam dn2.NDELAY = 4;
defparam dn3.NDELAY = 4;

// debugging
/*wire [26:0] x_sum,
wire [9:0] x_avg;
wire [26:0] new_x_avg;
wire [18:0] pixelCount;*/

```

```

        assign analyzer1_data = {pixelCount1[18:3]}; //{x_avg, pixelCount[18:16],
ntsc_we, startFinalAvgDividing, endFinalAvgDividing} ;
        assign analyzer1_clock = clk;
        assign analyzer2_data = {x1024, x1_avg[9:4]}; //{x, y[7:2]};
//vram_addr1[15:0];
        assign analyzer2_clock = clk;
        assign analyzer3_data = {pixelCount2[18:3]}; //{new_x_avg[9:0], y[1:0],
4'b0}; //{hcount[10:0], vcount[4:0]};
        assign analyzer3_clock = clk;
        assign analyzer4_data = {x720, x1_avg[3:0], 2'b0}; //{hcount[10:0], 5'b0};
//pixelCount[15:0];
        assign analyzer4_clock = clk;

        assign x720Prime=(x720>100) ? ((x720<512)? ((2*x720)-200):1024) : 0;
        assign y720Prime=y720;
Game
aGame(clock_65mhz,reset,hcount,vcount,hsync,vsync,blank,AICrosshairX,AICrosshairY,AIFire,x720Prime,y720Prime,filterWinner,~button3,~button2,~button1,

        phsync,pvsync,pblank,aPixel,duckX,duckY,duckAlive,AIShotsLeft,playerShotsLeft,roundOver,roundStart,dir,inCalibrate);

AI
aAI(clock_65mhz,reset,duckX,duckY,duckAlive,AIShotsLeft,roundStart,roundOver,dir,vsync,switch[7:6],
        AICrosshairX,AICrosshairY,AIFire);

        debounce aDebounce(reset,clock_65mhz,button_left,abutton_left);
        debounce bDebounce(reset,clock_65mhz,button_right,abutton_right);
        debounce cDebounce(reset,clock_65mhz,button_up,abutton_up);
        debounce dDebounce(reset,clock_65mhz,button_down,abutton_down);
        debounce eDebounce(reset,clock_65mhz,button3,aButton3);
        debounce fDebounce(reset,clock_65mhz,button2,aButton2);

        reg b,hs,vs;
        //playerInput
aPlayerInput(clock_65mhz,reset,~abutton_left,~abutton_right,~abutton_up,~abutton_down,~button0,playerShotsLeft,playerCrosshairX,playerCrosshairY,playerFire);
        always @ (posedge clock_65mhz) begin
                hs <= phsync;
                vs <= pvsync;
                b <= pblank;
                rgb <= aPixel;
        end

// VGA Output. In order to meet the setup and hold times of the
// AD7125, we send it ~clock_65mhz.
assign vga_out_red = inCalibrate ? pixel_out[2][23:16]:{rgb[7:5],5'b00000};
assign vga_out_green = inCalibrate ? pixel_out[2][15:8]:{rgb[4:2],5'b00000};
assign vga_out_blue = inCalibrate ? pixel_out[2][7:0]:{rgb[1:0],6'b000000};
assign vga_out_sync_b = 1'b1; // not used
assign vga_out_blank_b = inCalibrate ? ~b : ~bPrime;
assign vga_out_pixel_clock = ~clock_65mhz;
assign vga_out_hsync = inCalibrate ? hsPrime : hs;
assign vga_out_vsync = inCalibrate ? vsPrime : vs;

```



```
endmodule
```

XVGA Signal Generator

```
////////////////////////////////////  
/  
/  
//  
// xvga: Generate XVGA display signals (1024 x 768 @ 60Hz)  
//  
////////////////////////////////////  
/
```

```
module xvga(vclock,hcount,vcount,hsync,vsync,blank);  
    input vclock;  
    output [10:0] hcount;  
    output [9:0] vcount;  
    output vsync;  
    output hsync;  
    output blank;  
  
    reg hsync,vsync,hblank,vblank,blank;  
    reg [10:0] hcount; // pixel number on current line  
    reg [9:0] vcount; // line number  
  
    // horizontal: 1344 pixels total  
    // display 1024 pixels per line  
    wire hsynccon,hsyncoff,hreset,hblankon;  
    assign hblankon = (hcount == 1023);  
    assign hsynccon = (hcount == 1047);  
    assign hsyncoff = (hcount == 1183);  
    assign hreset = (hcount == 1343);  
  
    // vertical: 806 lines total  
    // display 768 lines  
    wire vsyncon,vsyncoff,vreset,vblankon;  
    assign vblankon = hreset & (vcount == 767);  
    assign vsyncon = hreset & (vcount == 776);  
    assign vsyncoff = hreset & (vcount == 782);  
    assign vreset = hreset & (vcount == 805);  
  
    // sync and blanking  
    wire next_hblank,next_vblank;  
    assign next_hblank = hreset ? 0 : hblankon ? 1 : hblank;  
    assign next_vblank = vreset ? 0 : vblankon ? 1 : vblank;  
    always @(posedge vclock) begin  
        hcount <= hreset ? 0 : hcount + 1;  
        hblank <= next_hblank;  
        hsync <= hsynccon ? 0 : hsyncoff ? 1 : hsync; // active low  
  
        vcount <= hreset ? (vreset ? 0 : vcount + 1) : vcount;  
        vblank <= next_vblank;  
        vsync <= vsyncon ? 0 : vsyncoff ? 1 : vsync; // active low  
  
        blank <= next_vblank | (next_hblank & ~hreset);  
    end  
endmodule
```

VRAM Display

```
////////////////////////////////////
```

```

// generate display pixels from reading the ZBT ram
// note that the ZBT ram has 2 cycles of read (and write) latency
//
// We take care of that by latching the data at an appropriate time.
//
// Note that the ZBT stores 36 bits per word; we use only 32 bits here,
// decoded into four bytes of pixel data.

module vram_display(reset,clk,hcount,vcount,vr_pixel,
                   vram_addr,vram_read_data);

    input reset, clk;
    input [10:0] hcount;
    input [9:0] vcount;
    output [23:0] vr_pixel;
    output [18:0] vram_addr;
    input [35:0] vram_read_data;

    wire [18:0] vram_addr;
    assign vram_addr = (vcount > 480 || hcount > 720) ?
        19'b0 :
        {vcount[8:0],
hcount[9:0]};

    wire [1:0] hc4 = hcount[1:0];

    assign vr_pixel = vram_read_data[23:0]; //Filtered RGB info!

    /*assign vr_pixel[1] = (hueFilterSwitch) ?
        (satpassHigh >
satpassLow ?
        (sat >
satpassLow && sat < satpassHigh ?
        vr_pixel[0] :
        hueFilterFailColor) :
        (sat >
satpassLow || sat < satpassHigh ?
        vr_pixel[0] :
        hueFilterFailColor) ) :
        vr_pixel[0]; */

endmodule // vram_display

```

Filter Limits Configuration

```
`timescale 1ns / 1ps
////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////
/
// Company:
// Engineer:
//
// Create Date:    11:16:06 11/30/06
// Design Name:
// Module Name:    limits_interface
// Project Name:
// Target Device:
// Tool versions:
// Description:
//
// Dependencies:
//
// Revision:
// Revision 0.01 - File Created
// Additional Comments:
//
////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////
/
module limits_interface(clk, reset, up_button, down_button, left_button,
right_button, enter_button,
                                hue1_lowMark,
hue1_highMark, hue2_lowMark, hue2_highMark,
                                sat1_lowMark,
sat1_highMark, sat2_lowMark, sat2_highMark, vfd_data);

    input clk, reset;
    input up_button, down_button, left_button, right_button, enter_button;
//raw button lines; assumes 1=pressed, 0=open

    output [8:0] hue1_lowMark, hue1_highMark, hue2_lowMark, hue2_highMark;
//Min and max values of the hue-pass, valued from 0 to 360
    output [6:0] sat1_lowMark, sat1_highMark, sat2_lowMark, sat2_highMark;
//Min and max values of the sat-pass, valued from 0 to 100 (0 == white, 100 ==
full color)

    output [16*8-1:0] vfd_data; //ascii output to vfd data (ie 16 8-bit
bytes, encoded as ASCII)

    //First lets synchronize and debounce the buttons
    wire up_button_s, down_button_s, left_button_s, right_button_s,
enter_button_s;
    wire up_button_d, down_button_d, left_button_d, right_button_d,
enter_button_d;
    synchronize sync1(clk, up_button,          up_button_s);
    synchronize sync2(clk, down_button,       down_button_s);
    synchronize sync3(clk, left_button,       left_button_s);
    synchronize sync4(clk, right_button,      right_button_s);
    synchronize sync5(clk, enter_button,      enter_button_s);

    debounce deb1(reset, clk, up_button_s,    up_button_d);
    debounce deb2(reset, clk, down_button_s,  down_button_d);
    debounce deb3(reset, clk, left_button_s,  left_button_d);
    debounce deb4(reset, clk, right_button_s, right_button_d);
```

```

debounce deb5(reset, clk, enter_button_s,      enter_button_d);

//Wire up the marker fsm's
wire doIndividualMove, doJointMove; //These wires go high for every clock
cycle where we move either

//and individual marker (selected by mode fsm) or both markers at once
slow_fast_fsm individual_fsm(reset, clk, (left_button_d ||
right_button_d), doIndividualMove);
//ie the left or right button control what speed we're moving the
individual marker

slow_fast_fsm      joint_fsm(reset, clk, (up_button_d || down_button_d),
doJointMove);
//ie the up or down buttons control what speed we're moving both
markers at once

//SELECTION: move between 0:MODE, 1:LOW, 2:HIGH, or 3:BOTH on left/right
buttons
parameter SELOBJ_MODE = 0;
parameter SELOBJ_LOW  = 1;
parameter SELOBJ_HI   = 2;
parameter SELOBJ_BOTH = 3;
reg [1:0] objectSelection = 0;

reg prev_right=0, prev_left = 0;
wire right_pulse = ~prev_right && right_button;
wire left_pulse  = ~prev_left  && left_button;
always @ (posedge clk) begin
    prev_right <= right_button;
    prev_left  <= left_button;

    if (right_pulse)
        objectSelection <= objectSelection + 1;
    else if (left_pulse)
        objectSelection <= objectSelection - 1;
end

//Wire up the limits controller
wire [8:0] hue1_lowMark, hue1_highMark, hue2_lowMark, hue2_highMark;
wire [6:0] sat1_lowMark, sat1_highMark, sat2_lowMark, sat2_highMark;
wire [1:0] modeSelection; //0:HUE1, 1:SAT1, 2:HUE2, 3:SAT2
limits_controller lc(reset, clk,
                    up_button_d,
down_button_d, left_button_d, right_button_d,
                    doIndividualMove,
doJointMove, objectSelection, modeSelection,
                    hue1_lowMark,
hue1_highMark, sat1_lowMark, sat1_highMark,
                    hue2_lowMark,
hue2_highMark, sat2_lowMark, sat2_highMark );

//Wire up the VFD controller
vfd_control vfdc(clk, reset,
                objectSelection, modeSelection,

```

```

sat1_lowMark, sat1_highMark,
sat2_lowMark, sat2_highMark,
hue1_lowMark, hue1_highMark,
hue2_lowMark, hue2_highMark,
vfd_data);

endmodule

module limits_controller(reset, clk,
                        up_button,
                        down_button, left_button, right_button,
                        doIndepMove,
                        doJointMove, objectSelection, modeSelection,
                        hue1_lowMark,
                        hue2_lowMark,
                        hue1_hiMark, sat1_lowMark, sat1_hiMark,
                        hue2_hiMark, sat2_lowMark, sat2_hiMark);

    input reset, clk, up_button, down_button, left_button, right_button;
    input doIndepMove, doJointMove; //Move the currently selected marker by
    itself or with its partner (ie high AND low together) -- merely tells which
    marker-control fsm the button came from

    input [1:0] objectSelection;
    parameter SELOBJ_MODE = 0;
    parameter SELOBJ_LOW = 1;
    parameter SELOBJ_HI = 2;
    parameter SELOBJ_BOTH = 3;

    output [1:0] modeSelection;
    reg [1:0] modeSelection = 0;
    parameter SELMODE_HUE1 = 0;
    parameter SELMODE_SAT1 = 1;
    parameter SELMODE_HUE2 = 2;
    parameter SELMODE_SAT2 = 3;

    output [8:0] hue1_lowMark, hue1_hiMark, hue2_lowMark, hue2_hiMark;
    //These get sent to the VFD display and to the ultimate output of the limits
    interface
        output [6:0] sat1_lowMark, sat1_hiMark, sat2_lowMark, sat2_hiMark;

    //Keep the individual marks as seperate modules
    mark the_hue1_hiMark( reset, clk, hue1_hiMark_incr, hue1_hiMark_decr,
hue1_hiMark);
    mark the_hue1_lowMark(reset, clk, hue1_lowMark_incr, hue1_lowMark_decr,
hue1_lowMark);
    defparam the_hue1_hiMark.START_VALUE = 35; //Add some arbitrary default
    values
    defparam the_hue1_lowMark.START_VALUE = 20;

    mark the_hue2_hiMark( reset, clk, hue2_hiMark_incr, hue2_hiMark_decr,
hue2_hiMark);
    mark the_hue2_lowMark(reset, clk, hue2_lowMark_incr, hue2_lowMark_decr,
hue2_lowMark);

```



```

        assign hue1_lowMark_incr= (modeSelection == SELMODE_HUE1 &&
//ie,
increment the low-mark when: we're adjusting hue1, AND
        (objectSelection == SELOBJ_LOW || objectSelection == SELOBJ_BOTH) &&
// we're adjusting either the LOW mark OR BOTH marks, AND
        up_button &&
doJointMove) ?
        // we pressed the up button AND the movementFSM
tells us to go
        1 : 0;

        assign hue1_hiMark_decr = (modeSelection == SELMODE_HUE1 &&
        (objectSelection == SELOBJ_HI || objectSelection == SELOBJ_BOTH) &&
        down_button
&& doJointMove) ?
        1 : 0;

        assign hue1_lowMark_decr= (modeSelection == SELMODE_HUE1 &&
        (objectSelection == SELOBJ_LOW || objectSelection == SELOBJ_BOTH) &&
        down_button
&& doJointMove) ?
        1 : 0;

//hue2
        assign hue2_hiMark_incr = (modeSelection == SELMODE_HUE2 &&
        (objectSelection == SELOBJ_HI || objectSelection == SELOBJ_BOTH) &&
        up_button &&
doJointMove) ?
        1 : 0;

        assign hue2_lowMark_incr= (modeSelection == SELMODE_HUE2 &&
        (objectSelection == SELOBJ_LOW || objectSelection == SELOBJ_BOTH) &&
        up_button &&
doJointMove) ?
        1 : 0;

        assign hue2_hiMark_decr = (modeSelection == SELMODE_HUE2 &&
        (objectSelection == SELOBJ_HI || objectSelection == SELOBJ_BOTH) &&
        down_button
&& doJointMove) ?
        1 : 0;

        assign hue2_lowMark_decr= (modeSelection == SELMODE_HUE2 &&
        (objectSelection == SELOBJ_LOW || objectSelection == SELOBJ_BOTH) &&

```

```

down_button
&& doJointMove) ?

1 : 0;

//saturation1
assign sat1_hiMark_incr = (modeSelection == SELMODE_SAT1 &&

(objectSelection == SELOBJ_HI || objectSelection == SELOBJ_BOTH) &&
up_button &&
doJointMove) ?

1 : 0;

assign sat1_lowMark_incr= (modeSelection == SELMODE_SAT1 &&

(objectSelection == SELOBJ_LOW || objectSelection == SELOBJ_BOTH) &&
up_button &&
doJointMove) ?

1 : 0;

assign sat1_hiMark_decr = (modeSelection == SELMODE_SAT1 &&

(objectSelection == SELOBJ_HI || objectSelection == SELOBJ_BOTH) &&
down_button
&& doJointMove) ?

1 : 0;

assign sat1_lowMark_decr= (modeSelection == SELMODE_SAT1 &&

(objectSelection == SELOBJ_LOW || objectSelection == SELOBJ_BOTH) &&
down_button
&& doJointMove) ?

1 : 0;

//hue2
assign sat2_hiMark_incr = (modeSelection == SELMODE_SAT2 &&

(objectSelection == SELOBJ_HI || objectSelection == SELOBJ_BOTH) &&
up_button &&
doJointMove) ?

1 : 0;

assign sat2_lowMark_incr= (modeSelection == SELMODE_SAT2 &&

(objectSelection == SELOBJ_LOW || objectSelection == SELOBJ_BOTH) &&
up_button &&
doJointMove) ?

1 : 0;

```



```

    assign sat2_hiMark_decr = (modeSelection == SELMODE_SAT2 &&

        (objectSelection == SELOBJ_HI || objectSelection == SELOBJ_BOTH) &&
        down_button
&& doJointMove) ?

                                                                    1 : 0;

    assign sat2_lowMark_decr= (modeSelection == SELMODE_SAT2 &&

        (objectSelection == SELOBJ_LOW || objectSelection == SELOBJ_BOTH) &&
        down_button
&& doJointMove) ?

                                                                    1 : 0;

endmodule

```

```

module blinker(clk, onOff);
    input clk;
    output onOff;

    reg onOff = 0;

    parameter MAXCOUNT = 25000000; //1/2 second blinking

    reg [25:0] count = 0;

    always @ (posedge clk) begin
        if (count == MAXCOUNT) begin
            count <= 0;
            onOff <= !onOff;
        end else
            count <= count + 1;
    end
endmodule

```

```

module vfd_control(clk, reset,

    objectSelection, modeSelection,
    hue1_low, hue1_high, sat1_low_in,
    sat1_high_in,
    hue2_low, hue2_high, sat2_low_in,
    sat2_high_in,
    vfd_data);

```

```

//Purpose: Output text "HUE LOWxxx HIxxx" or "SAT LOWxxx HIxxx" to VFD
module, blinking the mode thats appropriate
    input clk, reset;
    input [1:0] objectSelection, modeSelection;

```

```

parameter SELOBJ_MODE = 0;
parameter SELOBJ_LOW  = 1;
parameter SELOBJ_HI   = 2;
parameter SELOBJ_BOTH = 3;
parameter SELMODE_HUE1 = 0;
parameter SELMODE_SAT1 = 1;
parameter SELMODE_HUE2 = 2;
parameter SELMODE_SAT2 = 3;

input [8:0] hue1_low, hue1_high, hue2_low, hue2_high;
input [6:0] sat1_low_in, sat1_high_in, sat2_low_in, sat2_high_in;
wire [8:0] sat1_low = {2'b0, sat1_low_in}, //sign-extend saturation to
fit division format
                                sat1_high = {2'b0, sat1_high_in},
                                sat2_low  = {2'b0, sat2_low_in},
                                sat2_high = {2'b0, sat2_high_in};

output [8*16-1:0] vfd_data;

//Blink generator
wire blink;
blinker bl(clk, blink);

wire [8*4-1:0] textMode[1:0]; //partial string "H1 ",
"H2 ", "S1 ", "S2 " to be determined by mode and blinker
wire [8*3-1:0] textLow, textHi; //partial strings "LOW" and " HI"

assign textMode[0] = (modeSelection == SELMODE_HUE1) ?
(objectSelection == SELOBJ_MODE ? "H1 " : "h1 ") :
(modeSelection == SELMODE_SAT1 ?
(objectSelection == SELOBJ_MODE ? "S1 " : "s1 ") :
(modeSelection == SELMODE_HUE2 ?
(objectSelection == SELOBJ_MODE ? "H2 " : "h2 ") :
(objectSelection == SELOBJ_MODE ? "S2 " : "s2 ")
);

assign textMode[1] = (objectSelection == SELOBJ_MODE) ?
(blink ?
textMode[0] : " " ) :
(blink ?
textMode[0];

assign textLow = (objectSelection == SELOBJ_LOW || objectSelection ==
SELOBJ_BOTH) ? //"LOW" : "low";
(blink ? "LOW" : " " ) :
"low";
assign textHi  = (objectSelection == SELOBJ_HI || objectSelection ==
SELOBJ_BOTH) ? //" HI" : " hi";
(blink ? " HI" : " " ) :

```

```

        " hi";

//EXTRACTION OF 100's, 10's, and 1's DIGITS
// **ASSUMES max number will be 360**
wire [8:0] high, low; //hue high-mark and low-mark or sat high-mark and
low-mark, depending on the mode

assign high =modeSelection == SELMODE_HUE1 ?
    hue1_high :
    (modeSelection == SELMODE_SAT1 ?
        sat1_high :
        (modeSelection ==
SELMODE_HUE2 ?
            hue2_high :
            sat2_high));

assign low = modeSelection == SELMODE_HUE1 ?
    hue1_low :
    (modeSelection == SELMODE_SAT1 ?
        sat1_low :
        (modeSelection ==
SELMODE_HUE2 ?
            hue2_low :
            sat2_low));

reg [8:0] dividend;
reg [6:0] divisor; //will be either 100 or 10
wire [8:0] quot;
wire [6:0] remainder;
reg [6:0] lastRemainder;
wire ready;
divider3 div9x7(dividend, divisor, quot, remainder, clk, ready, 1'b0,
1'b0, 1'b0);
//Divider with dividend size 9 (511-0), divisor size 7 (127-0),
delay of 9+2 + 1=12 clock cycles

reg [7:0] highs_hundreds, lows_hundreds;
reg [7:0] highs_tens, highs_ones, lows_tens, lows_ones;

reg [1:0] hi_low = 0; //multiplex hi-val and low-val division
always @ (posedge clk) begin

    lastRemainder <= remainder;

    hi_low <= hi_low + 1;
    /* Divider has 12-cycle pipeline (ie get answer out 12
cycles later)
        when =0, send in hi / 100,
        get out hi/100 from 12 cycles ago
        =1, send in rem(hi/100) / 10, get
out rem(hi/100) /10 from 12 cycles ago
        =2, send in low / 100,
        get out low/100 from 12 cycles ago
        =3, send in rem(low/100)/ 10,
        get out rem(low/100)/10 from 12 cycles ago
        */

    case (hi_low)
        0: begin
            dividend <= high;
            divisor <= 100;

```

```

        highs_hundreds <= quot + 48;
    end
1: begin
    dividend <= lastRemainder; //remainder(hi/100)
    divisor <= 10;
    highs_tens <= quot + 48;
    highs_ones <= remainder + 48;
    end
2: begin
    dividend <= low;
    divisor <= 100;
    lows_hundreds <= quot + 48;
    end
3: begin
    dividend <= lastRemainder;
    divisor <= 10;
    lows_tens <= quot + 48;
    lows_ones <= remainder + 48;
    end
    endcase
end

wire [16*8-1:0] vfd_data = {textMode[1], textLow,

    lows_hundreds, lows_tens, lows_ones,

    textHi,

    highs_hundreds, highs_tens, highs_ones };

    /*
    ({6'b0, lows_hundreds} + 48), //+48 to get the number into the ASCII code
    (ie ASCII "0" == 0d48)

    ({4'b0, lows_tens} + 48),          //hopefully these all come out to be 8-
    bits wide

    ({4'b0, lows_ones} + 48),

    textHi,

    ({6'b0, highs_hundreds} + 48),    //+48 to get the number into the ASCII
    code (ie ASCII "0" == 0d48)

    ({4'b0, highs_tens} + 48),        //hopefully these all come out to
    be 8-bits wide

    ({4'b0, highs_ones} + 48) };*/

endmodule

```

```

module mark(reset, clk, increment, decrement, value);

```

```

//This module just keeps track of incrementing/decrementing a mark-value
and doing the wrap around.
//Mark value is 8-bits [0 and 360]

parameter MIN_VALUE = 0;
parameter MAX_VALUE = 360;

parameter START_VALUE = 5;

input reset, clk, increment, decrement; //We do +/- 1 each clock cycle
that increment and decrement are high (increment takes priority)
output [8:0] value;

reg [8:0] value = START_VALUE;

always @ (posedge clk) begin
    if (reset)
        value <= START_VALUE;
    else begin
        if (increment)
            if (value == MAX_VALUE)
                value <= MIN_VALUE;
            else
                value <= value + 1;
        else if (decrement)
            if (value == MIN_VALUE)
                value <= MAX_VALUE;
            else
                value <= value - 1;
        end
    end
end
endmodule

```

```

/*module selection_fsm(reset, next_button, prev_button, mode_select);

    input reset, next_button, prev_button;
    output [1:0] mode_select; //Says which parameter we're controlling...
0=hue, 1=low mark, 2=hi mark

    parameter MODE_HUE = 0;
    parameter MODE_LOWMARK = 1;
    parameter MODE_HIMARK = 2;

    reg [1:0] mode_select = MODE_HUE;

    always @ (posedge enter_button) begin //move only on posedge of the
enter_button
        if (reset)
            mode_select <= MODE_HUE;
        else begin
            if (mode_select == MODE_HIMARK)
                mode_select <= 0;

```

```

        else
            mode_select <= mode_select + 1;
        end
    end
endmodule
*/

```

```

module slow_fast_fsm(reset, clk, button, doMovePulse);

    input reset, clk, button; //button is the slow/fast button of interest
    output doMovePulse; //High for every clock cycle we want to increment the
value of interest

    parameter SLOW_LENGTH      = 130000000; //Num clock cycles to stay in
slow mode while button is depressed (max 256)
    parameter SLOW_TICK       = 30000000; //Num clock cycles between slow
doMovePulses, should be a factor of SLOW_INTERVAL (max 256)
    parameter FAST_TICK       = 010000000; //Num clock cycles between fast
doMovePulses (max 256)

    reg [26:0] count = 0, tick = 0;
    reg mode=0, doMovePulse;

    always @ (posedge clk) begin
        if (reset) begin
            count <= 0;
            mode <= 0;
        end else begin
            if (mode == 0) begin //SLOW MODE
                if (button) begin
                    if (count < SLOW_LENGTH) begin
                        count <= count + 1;

                        if (tick >= SLOW_TICK)
                            tick <= 0;
                        else
                            tick <= tick + 1;

                        if (tick == 0)
                            doMovePulse <= 1;
                        else begin
                            doMovePulse <= 0;
                        end

                    end else
                        mode <= 1; //Switch to fast mode

                end
            else begin //ie, if no button is pressed
                doMovePulse <= 0;
            end
        end
    end
endmodule

```

```

        count <= 0;
        tick <= 0;
    end

end //end slow mode

else begin //FAST MODE
    if (button) begin
        if (count >= FAST_TICK) begin
            count <= 0;
            doMovePulse <= 1;
        end else begin
            doMovePulse <= 0;
            count <= count + 1;
        end
    end else begin//ie, no button pressed
        doMovePulse <= 0;
        count <= 0;
        mode <= 0; //back to SLOW MODE;
    end
end

end

end
endmodule

```

NTSC TO ZBT And Filtering

```

//
// File:   ntsc2zbt.v
// Date:   27-Nov-05
// Author: I. Chuang <ichuang@mit.edu>
//
// Example for MIT 6.111 labkit showing how to prepare NTSC data
// (from Javier's decoder) to be loaded into the ZBT RAM for video
// display.
//
// The ZBT memory is 36 bits wide; we only use 32 bits of this, to
// store 4 bytes of black-and-white intensity data from the NTSC
// video input.

////////////////////////////////////
// Prepare data and address values to fill ZBT memory with NTSC data

module ntsc_to_zbt(clk, vclk, fvh, dv, din, ntsc_addr, ntsc_data, ntsc_we,
                  filterFail_bw_switch,
                  filterFail_color_switch,
                  hue1_passLow, hue1_passHigh,
                  hue2_passLow, hue2_passHigh,
                  sat1_passLow, sat1_passHigh,
                  sat2_passLow, sat2_passHigh,
                  hue1FilterSwitch,
                  sat1FilterSwitch, hue2FilterSwitch, sat2FilterSwitch,
                  weighted_x1_avg, weighted_x2_avg,
                  weighted_y1_avg, weighted_y2_avg, winningFilter,
                  noiseFilterSwitch,
                  filterDebounceSw,
                  pixelCount1, pixelCount2);
    //DEBUGS:
    //x_sum, new_x_avg, pixelCount,
startFinalAvgDividing, endFinalAvgDividing, x_out, y_out);

```

```

input      clk; // system clock
input      vclk; // video clock from camera
input [2:0] fvh;
input      dv;
input [29:0] din; // full ycrCb data
output [18:0] ntsc_addr;
output [35:0] ntsc_data; //in RGB!!
output      ntsc_we; // write enable for NTSC data
output [9:0] weighted_x1_avg, weighted_x2_avg; //x-average of all passing
pixels in filter1
output [9:0] weighted_y1_avg, weighted_y2_avg; //
                ..in filter2
output winningFilter; //(Debounced) Whether there is more in filter1
(==0) or filter2 (==1)
output [18:0] pixelCount1, pixelCount2;

input filterFail_bw_switch, filterFail_color_switch;
input [8:0] hue1_passLow, hue1_passHigh, hue2_passLow, hue2_passHigh;
input [6:0] sat1_passLow, sat1_passHigh, sat2_passLow, sat2_passHigh;
input hue1FilterSwitch, hue2FilterSwitch;
input sat1FilterSwitch, sat2FilterSwitch;
input noiseFilterSwitch;
input filterDebounceSw;

//DEBUGS:
/*output [26:0] x_sum;
output [26:0] new_x_avg;
output [18:0] pixelCount;
output startFinalAvgDividing, endFinalAvgDividing;
output [9:0] x_out;
output [7:0] y_out; */

wire [23:0] vr_pixel; //This is the output pixel line.
wire [23:0] vr_pixel_hue1pass, vr_pixel_hue2pass;
wire [23:0] vr_pixel_sat1pass, vr_pixel_sat2pass;

reg [9:0] col = 0;
reg [7:0] row = 0;
reg [29:0] vdata = 0;
reg vwe;
reg old_dv;
reg old_frame; // frames are even / odd interlaced
reg even_odd; // decode interlaced frame to this wire

wire frame = fvh[2];
wire frame_edge = frame & ~old_frame;

//*****
// START COLOR-SPACE CONVERTERS
// all color space converters use 65mhz clock

```



```

////////////////////////////////////
// YCrCb->RGB Color Space Converter (with Y Cr Cb as 10-bit inputs, R G B
as 8-bit outputs, 3 clock cycle delay */
wire [9:0] Ylum, Cr, Cb;
wire [7:0] R, G, B;
YCrCb2RGB ycrCb_csc( R, G, B, clk, reset, Ylum, Cr, Cb );

////////////////////////////////////
//RGB->Hue COLOR SPACE CONVERTER (24 cycle delay)
wire [8:0] hue; //between 0 and 360
wire [6:0] sat;
RGB2Hue rgb_csc(clk, reset, R, G, B, hue, sat);

//Still want to use RGB later, so need to delay it by 24 cycles to match
it with the hue output
wire [7:0] Rd, Gd, Bd;
parameter RGB_delay = 24;
delayNxM Rder(clk, R, Rd);
delayNxM Gder(clk, G, Gd);
delayNxM Bder(clk, B, Bd);
defparam Rder.MSIZE = 8;
    defparam Rder.dn[7].NDELAY = RGB_delay;
    defparam Rder.dn[6].NDELAY = RGB_delay;
    defparam Rder.dn[5].NDELAY = RGB_delay;
    defparam Rder.dn[4].NDELAY = RGB_delay;
    defparam Rder.dn[3].NDELAY = RGB_delay;
    defparam Rder.dn[2].NDELAY = RGB_delay;
    defparam Rder.dn[1].NDELAY = RGB_delay;
    defparam Rder.dn[0].NDELAY = RGB_delay;
defparam Gder.MSIZE = 8;
    defparam Gder.dn[7].NDELAY = RGB_delay;
    defparam Gder.dn[6].NDELAY = RGB_delay;
    defparam Gder.dn[5].NDELAY = RGB_delay;
    defparam Gder.dn[4].NDELAY = RGB_delay;
    defparam Gder.dn[3].NDELAY = RGB_delay;
    defparam Gder.dn[2].NDELAY = RGB_delay;
    defparam Gder.dn[1].NDELAY = RGB_delay;
    defparam Gder.dn[0].NDELAY = RGB_delay;
defparam Bder.MSIZE = 8;
    defparam Bder.dn[7].NDELAY = RGB_delay;
    defparam Bder.dn[6].NDELAY = RGB_delay;
    defparam Bder.dn[5].NDELAY = RGB_delay;
    defparam Bder.dn[4].NDELAY = RGB_delay;
    defparam Bder.dn[3].NDELAY = RGB_delay;
    defparam Bder.dn[2].NDELAY = RGB_delay;
    defparam Bder.dn[1].NDELAY = RGB_delay;
    defparam Bder.dn[0].NDELAY = RGB_delay;

//If displaying just B&W data, want to delay it by 27 cycles so it still
stays sync'd with RGB mode
wire [7:0] BW_delayed;
parameter BW_delay = 27; //3 for RGB + 24 for hue
delayNxM BWder(clk, Ylum, BW_delayed);
defparam BWder.MSIZE = 8;
    defparam BWder.dn[7].NDELAY = BW_delay;
    defparam BWder.dn[6].NDELAY = BW_delay;
    defparam BWder.dn[5].NDELAY = BW_delay;
    defparam BWder.dn[4].NDELAY = BW_delay;
    defparam BWder.dn[3].NDELAY = BW_delay;
    defparam BWder.dn[2].NDELAY = BW_delay;
    defparam BWder.dn[1].NDELAY = BW_delay;

```

```

        defparam BWder.dn[0].NDELAY = BW_delay;

//END COLOR-SPACE CONVERTERS
//*****

//Address Delay -- Synchronize generated address with the results of hue
and sat filtering
wire [18:0] myaddr;
wire [18:0] ntsc_addr;
parameter addr_delay = BW_delay; //3 for RGB + 24 for hue
delayNxM addr_der(clk, myaddr, ntsc_addr);
defparam addr_der.MSIZE = 19;

        defparam addr_der.dn[18].NDELAY = addr_delay;
        defparam addr_der.dn[17].NDELAY = addr_delay;
        defparam addr_der.dn[16].NDELAY = addr_delay;
        defparam addr_der.dn[15].NDELAY = addr_delay;
        defparam addr_der.dn[14].NDELAY = addr_delay;
        defparam addr_der.dn[13].NDELAY = addr_delay;
        defparam addr_der.dn[12].NDELAY = addr_delay;
        defparam addr_der.dn[11].NDELAY = addr_delay;
        defparam addr_der.dn[10].NDELAY = addr_delay;
        defparam addr_der.dn[9].NDELAY = addr_delay;
        defparam addr_der.dn[8].NDELAY = addr_delay;
        defparam addr_der.dn[7].NDELAY = addr_delay;
        defparam addr_der.dn[6].NDELAY = addr_delay;
        defparam addr_der.dn[5].NDELAY = addr_delay;
        defparam addr_der.dn[4].NDELAY = addr_delay;
        defparam addr_der.dn[3].NDELAY = addr_delay;
        defparam addr_der.dn[2].NDELAY = addr_delay;
        defparam addr_der.dn[1].NDELAY = addr_delay;
        defparam addr_der.dn[0].NDELAY = addr_delay;

//////////
//VIDEO READING STUFF

always @ (posedge vclk) //LLC1 is reference
begin
    old_dv <= dv;
    vwe <= dv && !fvh[2] & ~old_dv; // if data valid, write it
    old_frame <= frame;
    even_odd = frame_edge ? ~even_odd : even_odd;

    if (!fvh[2])
        begin
            col <= fvh[0] ? 0 : //on a horizontal-sync, reset column
                (!fvh[2] && !fvh[1] && dv ) ? col + 1 : col; //should
effectively go up to 720
            row <= fvh[1] ? 0 : //on a vertical-sync, reset row
                (!fvh[2] && fvh[0] ) ? row + 1 : row;
//should effectively go up to 480/2 (interlacing)
            vdata <= (dv && !fvh[2]) ? din : vdata;
        end
end

// synchronize with system clock      (above was sync'd with
tv_in_line_clock1)

reg [9:0] x[1:0]; //two-register synchronizers
reg [7:0] y[1:0]; //''
reg [29:0] data[1:0]; //''

```

```

reg      we[1:0];          //'
reg      eo[1:0];         //'

always @(posedge clk)
begin
    {x[1],x[0]} <= {x[0],col};
    {y[1],y[0]} <= {y[0],row};
    {data[1],data[0]} <= {data[0],vdata};
    {we[1],we[0]} <= {we[0],vwe};
    {eo[1],eo[0]} <= {eo[0],even_odd};

end

//WRITE-ENABLE GENERATION
// edge detection on write enable signal
reg old_we;
wire we_edge = we[1] & ~old_we;
always @(posedge clk) old_we <= we[1];

//STORAGE ADDRESS COMPUTATION
//Address to store in ZBT-ram is basically the y-location plus the x-
location

assign myaddr = {y[1][7:0], eo[1], x[1][9:0]};
//wire ntsc_addr (output) is my_addr delayed by 3+24 @ 65mhz (ie its the
myaddr of the output of the hue and sat filters)

delayN we_der(clk, we_edge, ntsc_we);
defparam we_der.NDELAY = addr_delay; //Delay the generated we by 3+24 to
synchronize it with outputs of filters

//COLOR SPACE CONVERSION
// YCrCb->RGB Color Space Converter (with Y Cr Cb as 10-bit inputs, R G B
as 8-bit outputs, 3 clock cycle delay */
assign Ylum = data[1][29:20] ; //luminance data
assign Cr = data[1][19:10] ; //Cr data
assign Cb = data[1][ 9: 0] ; //Cb data

/*WIRE LIST AT THIS POINT:
-----
R, G, B are Ylum, Cr, and Cb converted and delayed by 3 @ 65mhz
hue and sat are R, G, B converted and delayed by 24 @
65mhz
Rd, Gd, Bd are R, G, B
delayed by 24 @ 65mhz
BW_delayed is Ylum
delayed by 3+24 @ 65mhz */

```

```

/*OUTPUT PIXEL GENERATION:
    Output to ZBT ram will be in RGB format.
    To create output pixel, we want to send it first through two
filters.

    First we send it through a hue-pass filter.  If the pixel's hue is
inside the specified range
    of allowable hues, we pass the RGB information on to the next
filter.

    The next filter is a saturation-pass filter.  If the pixel's
saturation is inside the specified range
    of allowable saturations, we pass the RGB information out to the
memory.

    If the pixel fails any of these filters, we pass through a "fail-
color" to be written to ZBT memory
    for that particular location.    */

/*Fail-Color:
    If we fail the hue and sat limits filter, we have 3 options:
    if bw_switch: display failed pixels as b&w
    if colorFail_switch: display failed pixels as pure blue
    if neither: display failed pixels as black */
wire [24:0] hueFilterFailColor = filterFail_bw_switch ?

{BW_delayed,BW_delayed,BW_delayed} :

(filterFail_color_switch ? {8'b0, 8'b0, 8'd255} : 24'b0);

wire pixel_failed_huesatFilter1, pixel_failed_huesatFilter2;

hueSatFiltering huesat1_filter( hue1FilterSwitch, sat1FilterSwitch,
hue, sat,
hue1_passHigh, hue1_passLow, sat1_passHigh, sat1_passLow,
hue1_Failed, sat1_Failed, pixel_failed_huesatFilter1);

hueSatFiltering huesat2_filter( hue2FilterSwitch, sat2FilterSwitch,
hue, sat,
hue2_passHigh, hue2_passLow, sat2_passHigh, sat2_passLow,
hue2_Failed, sat2_Failed, pixel_failed_huesatFilter2);

wire pixel_failed_huesat = pixel_failed_huesatFilter1 &&
pixel_failed_huesatFilter2; //Did the pixel fail both enabled filters (ie
passes either filter)?

/*NOISE FILTER:
    For a pixel to be displayed and count, the last
number_of_past_passing_pixels pixels have to
    have passed the hue and sat filters*/
reg    [6:0] past_passingFilter1_pixel_buffer = 0;
reg    [6:0] past_passingFilter2_pixel_buffer = 0;

```

```

parameter LAST_7_PIXELS = 7'b1111111;

always @ (posedge clk) begin
    if (x[1] < 2) begin //Reset pixel buffer at the start of a row
        past_passingFilter1_pixel_buffer <= 0;
        past_passingFilter2_pixel_buffer <= 0;
    end else begin
        past_passingFilter1_pixel_buffer <=
{past_passingFilter1_pixel_buffer[5:0], ~pixel_failed_huesatFilter1};
        past_passingFilter2_pixel_buffer <=
{past_passingFilter2_pixel_buffer[5:0], ~pixel_failed_huesatFilter2};
    end
end

wire pixel_passed_noiseFilter1, pixel_passed_noiseFilter2;
assign pixel_passed_noiseFilter1 = past_passingFilter1_pixel_buffer[6:0]
== LAST_7_PIXELS;
assign pixel_passed_noiseFilter2 = past_passingFilter2_pixel_buffer[6:0]
== LAST_7_PIXELS;

//Output Pixel:
wire pixelPassedFilters1 = noiseFilterSwitch ?

pixel_passed_noiseFilter1 : //NOTE: Implicit that if it passed the
noise filter, it passed the hue-sat filter as well

~pixel_failed_huesatFilter1;
wire pixelPassedFilters2 = noiseFilterSwitch ?

pixel_passed_noiseFilter2 :

~pixel_failed_huesatFilter2;
//Note: If filter1 (ie hue1 and sat1) or filter2 (ie hue2 and sat2) is
disabled,
// the pixel "fails" the appropriate filter

assign vr_pixel = (pixelPassedFilters1 || pixelPassedFilters2) ? //If
the pixel passes either noise filter...

{Rd,Gd,Bd} :
//display it as rgb
hueFilterFailColor;

//Else, kill it

wire [35:0] ntsc_data;
assign ntsc_data = {12'b0, vr_pixel};

//CENTROID CALCULATOR

```

```

wire [9:0] x1_avg, x2_avg, y1_avg, y2_avg;
wire [18:0] pixelCount1, pixelCount2;

centroid_calculator filter1_centroid(clk, x[1], y[1], eo[1], ntsc_we,
~pixelPassedFilters1, x1_avg, y1_avg, pixelCount1);
centroid_calculator filter2_centroid(clk, x[1], y[1], eo[1], ntsc_we,
~pixelPassedFilters2, x2_avg, y2_avg, pixelCount2);

weighted_averager w_avger1(clk, x1_avg, (x[1]==0 && y[1]==0),
weighted_x1_avg);
weighted_averager w_avger2(clk, x2_avg, (x[1]==0 && y[1]==0),
weighted_x2_avg);
weighted_averager w_avger3(clk, y1_avg, (x[1]==0 && y[1]==0),
weighted_y1_avg);
weighted_averager w_avger4(clk, y2_avg, (x[1]==0 && y[1]==0),
weighted_y2_avg);

//OUTPUT DEBOUNCING
wire noisyWinner; //The filter that has most pixels detected at the
moment (0 == filter1, 1==filter2)
assign noisyWinner = (pixelCount2 > pixelCount1) ? 1:0;

reg newWinner = 0; //Which filter we are suspecting to be the winner (0
== filter1, 1==filter2)
reg cleanFilterWinner = 0; //Debounced winning filter

reg [4:0] debounceCount=0; //Number of NTSC frames the current winning
filter has been the winning filter
parameter FILTER_DEBOUNCE_TIME = 5; //Number of NTSC frames to a filter
needs to be winning before its considered the winner (note: 1 frame = 37ms)

always @ (posedge clk) begin
    if (x[1] == 0 && y[1] == 0) begin //ie check every new frame
        if (noisyWinner != newWinner) begin
            newWinner <= noisyWinner;
            debounceCount <= 0;
        end else if (debounceCount == FILTER_DEBOUNCE_TIME)
            cleanFilterWinner <= newWinner;
        else
            debounceCount <= debounceCount+1;
    end
end

assign winningFilter = filterDebounceSw ?
cleanFilterWinner :
noisyWinner;

endmodule // ntsc_to_zbt

module hueSatFiltering(hueFilterSwitch, satFilterSwitch,
hue, sat,
huepassHigh, huepassLow,
satpassHigh, satpassLow,
hueFailed, satFailed,
pixelFailed);

/*Hue/Sat-Pass Filter:
hue1_passLow, hue1_passHigh, hue2_passLow, hue2_passHigh,

```

```

        sat1_passLow, sat1_passHigh, sat2_passLow, sat2_passHigh,

        If huepassHigh > huepassLow, pass any pixel between low and high
        If huepassHigh < huepassLow, pass any pixel not between low and
high    */

        input hueFilterSwitch, satFilterSwitch;
        input [8:0] hue, huepassHigh, huepassLow;
        input [6:0] sat, satpassHigh, satpassLow;
        output hueFailed, satFailed, pixelFailed;

        assign hueFailed = (hueFilterSwitch) ?
?
        (huepassHigh > huepassLow
        (hue > huepassLow
        0 :
        1) :
        (hue > huepassLow
        0 :
        1) ) :
        0;

        assign satFailed = (satFilterSwitch) ?
        (satpassHigh >
        (sat >
        0 :
        1) :
        (sat >
        0:
        1) ) :
        0;

        satpassLow ?
        (sat >
        0 :
        1) :
        (sat >
        0:
        1) ) :
        0;

        satpassLow && sat < satpassHigh ?
        (sat >
        0 :
        1) :
        (sat >
        0:
        1) ) :
        0;

        satpassLow || sat < satpassHigh ?
        0;

        /*For pixelFailed, we want:
        If both filters are enabled, fail (=1) if it fails either filter.
        If one filter is enabled, fail (=1) if it fails just that
filter
        If both filters are disabled, fail (=1) always . */

        assign pixelFailed = (hueFilterSwitch || satFilterSwitch) ?
        ((hueFilterSwitch
        && hueFailed) || (satFilterSwitch && satFailed)) :
        1;
        //Thus if both hue and sat filters are disabled, the pixel
        "fails" the filter.

        satFilterSwitch) ?
        (hueFailed ||
        (hueFilterSwitch ?
        hueFailed :
        (satFilterSwitch ?

```

```

        satFailed :
                                                    0));*/
endmodule

module centroid_calculator(clk, x, y, eo, ntsc_we, pixelFailed, x_avg, y_avg,
pixelCount_out);

    input          clk;
    input [9:0]    x;
    input [7:0]    y;
    input          eo;
    input          ntsc_we;
    input          pixelFailed;
    output [9:0]   x_avg, y_avg;
    output [18:0] pixelCount_out;

/*Averages/Centroid Calculation
    Centroid calculation is done by simply averaging all passing
pixels. But what size should
    the sum register be?
    Worst-case scenario: every pixel in a row passes. The max value
of the row's x-sum
    is then:
         $720/2 * 720 + 720/2 = 259,560$  (proving this is left as an
exercice to the reader).
    Assuming every pixel in every row passes, the max value of the
total x-sum is:
         $259,560 * 480 = 124,588,800$ 
 $2^{27}$  is:  $134,217,128$ 
    so lets make the x-sum variable 27-bits long.

    What size should the count register be? Worst case scenario is:
         $480 * 720 = 345,600$ 
 $2^{19}$  is  $524,288$ 
*/
    reg [26:0] x_sum = 0, y_sum=0;
    wire [26:0] new_x_sum, new_y_sum;
    wire [9:0] new_pixel_x, new_pixel_y;
    reg latch_x_sum, latch_y_sum;

    reg [9:0] x_avg = 0, y_avg=0;

    reg [18:0] pixelCount = 0;
    reg [18:0] pixelCount_out = 0;

    xy_adder2 x_summer(x_sum, new_pixel_x, new_x_sum, clk);
    xy_adder2 y_summer(y_sum, new_pixel_y, new_y_sum, clk);

    wire [26:0] new_x_avg, new_y_avg;
    wire startFinalAvgDividing, endFinalAvgDividing;
    delayN validAvgDivision(clk, startFinalAvgDividing, endFinalAvgDividing);
    defparam validAvgDivision.NDELAY = 30; //Divider's delay is 29 clocks.
start is high when the final values go in, so end will be high when final avg
comes out

    xyavg_divider x_avger(new_x_sum, pixelCount, new_x_avg, junk1, clk,
junk2, junk3, junk4, junk5);

```



```

        xyavg_divider y_avger(new_y_sum, pixelCount, new_y_avg, junk6, clk,
junk7, junk8, junk9, junk10);

/*CENTROID CALCULATION */

        assign new_pixel_x = pixelFailed ? 0 : x; //Add the current pixel's x-
value to the sum only when it passes the filters
        assign new_pixel_y = pixelFailed ? 0 : {1'b0, y, eo};

        assign startFinalAvgDividing = (x == 718 && y == 238) ? 1 : 0; //y==238
because y is interlaced, so y=238 corresponds to line 476 or 477, depending on
eo[0]
        //endFinalAvgDividing is startFinalAvgDividing delayed by 29+1

        always @ (posedge clk) begin
            latch_x_sum <= ntsc_we; //ntsc_we is valid when the output pixel
and address info is correct.
                                                    //Since the
adder is adding passing pixels and has a latency of 1, the cycle
                                                    //after
ntsc_we is high is when we have a new valid sum.

            if (x < 1 && y < 1) begin
                //If we're starting a new frame, reset all the averages
                //x_avg <= 0;
                //y_avg <= 0;
                x_sum <= 0;
                y_sum <= 0;

                pixelCount <= 0;
            end else begin

                //Update sums
                if (latch_x_sum) begin
                    x_sum <= new_x_sum;
                    y_sum <= new_y_sum;
                end

                //Update the pixel count
                if (ntsc_we && !pixelFailed)
                    pixelCount <= pixelCount + 1;

                if (startFinalAvgDividing)
                    pixelCount_out <= pixelCount;

                //Latch the output average to the calculated average from
the (pipelined) divider
                else if (endFinalAvgDividing) begin //Output of divider
right now is the final x_avg
                    x_avg <= new_x_avg[9:0];
                    y_avg <= new_y_avg[9:0];
                end
            end
        end
endmodule

```

```

module weighted_averager(clk, new_input, new_data_enable, prev_average);
    /* DOES A TIME-WEIGHTED AVERAGE:

```

basically

$$\frac{x_0}{2} + \frac{x_1}{4} + \frac{x_2}{6} + \frac{x_3}{8} + \frac{x_4}{10} + \dots$$

```

        if x_4 is oldest input and x_0 is current input, output is
    input clk;
    input [10:0] new_input;
    input new_data_enable;
    output [10:0] prev_average;

    reg [11:0] prev_average = 0;
    reg [10:0] new_input_latched = 0;
    wire [10:0] weighted_average = prev_average[10:0];
    wire [11:0] new_weighted_average;
    weighted_average_divider w_avg_diver(prev_average + new_input_latched,
3'd2, new_weighted_average,

    junk1, clk, junk2, junk3, junk4, junk5);

    always @ (posedge clk) begin
        if (new_data_enable) begin
            new_input_latched <= new_input;
            prev_average <= new_weighted_average;
        end
    end
endmodule

```

RGB To HSV Color Converter

```

module delayNxM(clk,in,out);

    //parameter NDELAY = 3; //Number of cycles to delay, min=2
    parameter MSIZE = 1; //Number of bits to delay (ie 2 == input is [1:0]),
min=1

    input clk;
    input [MSIZE-1:0] in;
    output [MSIZE-1:0] out;

    delayN dn[MSIZE-1:0] ( {MSIZE{clk}}, in, out);
    //defparam dn.NDELAY = NDELAY;

endmodule // delayN

module RGB2Hue(clk, reset, Rin, Gin, Bin, hueOut, satBuffered );

    input [7:0] Rin, Gin, Bin;
    input clk,reset;

    wire [8:0] R, G, B; //We need to do arithmitic with signed variables,
so here
    assign R = {1'b0, Rin}; //we're just sign-extending RGB
    assign G = {1'b0, Gin};
    assign B = {1'b0, Bin};

```

```

output [8:0] hueOut; //hue will be value between 0 and 360
output [6:0] satBuffered; //saturation will be value between 0 and 100

//HUE PIPELINE REGISTERS AND WIRES
//////////

//STAGE 1:
reg signed [8:0] Rd=0, Gd=0, Bd=0; //signed RGB delayed by 1
reg signed [8:0] max=0, min=0; //Max and min RGB values

reg [2:0] mode = 0; //notes which value was the maximum
parameter MAXeqMIN = 0 ;
parameter MAXisR_GgB = 1;
parameter MAXisR_GlB = 2;
parameter MAXisG = 3;
parameter MAXisB = 4;

wire [8:0] num_a, num_b; //These go into numerator subtractor.
Combinational logic selects which RGB values go into these.

//STAGE 2:
wire [8:0] denominator; //Denom[0] is used in stage 2, then delayed 2
cycles while the numerator is being multiplied
wire [8:0] denominatorD;
wire [8:0] numerator;

//STAGE 3:
wire [2:0] modeD; //mode is used in stage 1, then gets delayed 22
cycles to middle of stage 3
wire [14:0] numeratorX60; //Numerator from STAGE 2 multiplied by 6
wire hueUndefined;

//STAGE 4:
wire [14:0] quotient;
wire [8:0] addTerm;

//STAGE 5:
wire [16:0] hue; //Hue before checking for undefined (gray) condition
wire [8:0] hueOut;
reg hueUndefinedD; //hueUndefined delayed by 2 cycles

//STAGE 6:
reg [8:0] hueBuffered = 0; //adder takes too long to be accurate. Make
the output a buffered register.

//SAT PIPELINE REGISTERS AND WIRES
//////////
wire [14:0] minX100;
wire [7:0] maxD;
wire [14:0] satDivisionResult;
wire [6:0] satSubtractionResult;
wire [6:0] satChecked;
wire [6:0] satOut;
reg [6:0] satBuffered;

//PIPELINE DELAYS
//////////

```

```

//HUE
////
//delayN d1[2:0];
delayNxM d1(clk, mode, modeD); //Which formula we're using (to select
which number we add in stage 3
defparam d1.MSIZE = 3;
defparam d1.dn[2].NDELAY = 22;
defparam d1.dn[1].NDELAY = 22;
defparam d1.dn[0].NDELAY = 22;

delayNxM d2(clk, denominator, denominatorD);
defparam d2.MSIZE = 9;
defparam d2.dn[8].NDELAY = 2;
defparam d2.dn[7].NDELAY = 2;
defparam d2.dn[6].NDELAY = 2;
defparam d2.dn[5].NDELAY = 2;
defparam d2.dn[4].NDELAY = 2;
defparam d2.dn[3].NDELAY = 2;
defparam d2.dn[2].NDELAY = 2;
defparam d2.dn[1].NDELAY = 2;
defparam d2.dn[0].NDELAY = 2;

//delayN d3(clk, hueUndefined, hueUndefinedD);
//defparam d3.NDELAY = 1;
//Note: can't delay by 1 using NDELAY. So just doing it in sequential
area.

//SAT
////
delayNxM d4(clk, max[7:0], maxD);
defparam d4.MSIZE = 8;
defparam d4.dn[7].NDELAY = 2;
defparam d4.dn[6].NDELAY = 2;
defparam d4.dn[5].NDELAY = 2;
defparam d4.dn[4].NDELAY = 2;
defparam d4.dn[3].NDELAY = 2;
defparam d4.dn[2].NDELAY = 2;
defparam d4.dn[1].NDELAY = 2;
defparam d4.dn[0].NDELAY = 2;

delayNxM d5(clk, satChecked, satOut);
defparam d5.MSIZE = 7;
defparam d5.dn[6].NDELAY = 3;
defparam d5.dn[5].NDELAY = 3;
defparam d5.dn[4].NDELAY = 3;
defparam d5.dn[3].NDELAY = 3;
defparam d5.dn[2].NDELAY = 3;
defparam d5.dn[1].NDELAY = 3;
defparam d5.dn[0].NDELAY = 3;

//ANY COMBINATIONAL LOGIC
////////////////////////////////////
assign num_a = (mode == MAXeqMIN ?
MAXisR_GlB ?
8'b0 :
(mode == MAXisR_GgB || mode ==
Gd :
(mode == MAXisG ?

```

```

Bd :
Rd) ) );

assign num_b = (mode == MAXeqMIN ?
MAXisR_GlB ?
8'b0 :
(mode == MAXisR_GgB || mode ==
Bd :
(mode == MAXisG ?
Rd :
Gd) ) );

assign addTerm = (modeD == MAXisR_GlB ?
9'd360 :
(modeD == MAXisG ?
9'd120 :
(modeD == MAXisB ?
9'd240 :
9'b0) ) );

assign hueUndefined = (modeD == MAXeqMIN);
assign hueOut = hueUndefinedD ? //If the hue is undefined, we
want to ultimately output 0.
9'b0 : hue[8:0]; //Otherwise, we
want to output the low 8 bits of the adder

assign satChecked = (satSubtractionResult > 100) ? //Rounding bugs
in subtractor sometimes give us values more than 100. Don't let that.
7'd100 :
satSubtractionResult;

//WIRE UP ALL THE MATH-ERS
////////////////////////////////////
//HUE
hsv_subtractor sub1(max, min, denominator, clk); //Denominator <= max-
min
hsv_subtractor sub2(num_a, num_b, numerator, clk);

hsv_multiplier mult1(clk, numerator, numeratorX60);

hsv_divider div1(numeratorX60, denominatorD, quotient, junk1, clk,
junk2, junk3, junk4, junk5);

hsv_adder add1(quotient, addTerm, hue, clk);

//SAT
hsv_s_multiplier mult2(clk, min[7:0], minX100); //min[7:0] b/c min was
designed to be signed, so it's actually 8. Its always positive, though, so
drop the sign

hsv_s_divider div2(minX100, maxD, satDivisionResult, junk6, clk,
junk7, junk8, junk9, junk10);

wire [6:0] hsv_subtractor_a = 7'd100;

```

```

    hsv_s_subtractor sub3(hsv_subtractor_a, satDivisionResult[6:0],
satSubtractionResult, clk); //We set it up so division result is no more than
100, so use only [6:0]

```

```

always @ (posedge clk) begin

```

```

////////////////////////////////////
//STAGE 1: MIN/MAX SELECTOR and RGB DELAY
//Length: 1 clock cycle

```

```

//First step: sort out the min and max values

```

```

//Red is max or max=min

```

```

if (R >= G && R >= B) begin

```

```

    max <= R;

```

```

    if (G > B) begin

```

```

        min <= B;

```

```

        if (R == B)

```

```

            mode <= MAXeqMIN;

```

```

        else

```

```

            mode <= MAXisR_GgB; //...and green is

```

```

greater than blue

```

```

    end else begin

```

```

        min <= G;

```

```

        if (R == G)

```

```

            mode <= MAXeqMIN;

```

```

        else

```

```

            mode <= MAXisR_GlB; //...and green is less

```

```

than blue

```

```

    end

```

```

//Green is max

```

```

end else if (G > R && G > B) begin

```

```

    mode <= MAXisG;

```

```

    max <= G;

```

```

    if (R > B)

```

```

        min <= B;

```

```

    else

```

```

        min <= R;

```

```

//Blue is max

```

```

end else begin

```

```

    mode <= MAXisB;

```

```

    max <= B;

```

```

    if (R > G)

```

```

        min <= G;

```

```

    else

```

```

        min <= R;

```

```

end

```

```

//Next delay R, G, B to fit with pipeline and get ready for signed
arithmitic // hopefully R = 100 will be Rd = +100
Rd <= R;
Gd <= G;
Bd <= B;

////////////////////////////////////
//STAGE 2: NUMERATOR AND DENOMONATOR GENERATION
// Latency: 1 (a 9-bit signed subtractor w/ 9-bit signed
output)

////////////////////////////////////
//STAGE 3: MULTIPLIER
// Multiply numerator by 0d60
// Latency: 2 (9-bit signed multiplier by constant 0d60,
output is 15-bit signed)

////////////////////////////////////
//STAGE 4: DIVIDER and ADDING SELECTOR
// Divide the numerator*60 by denominator
// Latency: 15+4=19 (15-bit-signed-numerator divider)

hueUndefinedD <= hueUndefined;

hueBuffered <= hueOut;
satBuffered <= satOut;

end

endmodule

```

```

////////////////////////////////////
//
// Pushbutton Debounce Module (video version)
//
////////////////////////////////////

module debounce (reset, clock_65mhz, noisy, clean);
input reset, clock_65mhz, noisy;
output clean;

reg [19:0] count;

```

```

reg new, clean;

always @(posedge clock_65mhz)
    if (reset) begin new <= noisy; clean <= noisy; count <= 0; end
    else if (noisy != new) begin new <= noisy; count <= 0; end
    else if (count == 650000) clean <= new;
    else count <= count+1;

endmodule

```

Game Engine

```

////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////
//
// Game: game engine
//
////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////

module Game
(clock, reset, hCount, vCount, hsync, vsync, blank, AICrosshairX, AICrosshairY, AIFire, p
layerCrosshairX, playerCrosshairY, playerFire, calibrate, startPlaying, leaveCalibra
te, phsync, pvsync, pblank, pixel, duckX, duckY, duckAlive, AIShotsLeft, playerShotsLeft
, roundOver, roundStart, dir, inCalibrate);
    input clock; //system clock
    input reset; //global reset, active high
    input [10:0] hCount; //horizontal position of the pixel
being requested;
    input [9:0] vCount; //vertical position of the pixel being
requested;
    input hsync, vsync; //horizontal and vertical sync, active low
    input blank; //xvga blanking, 1 means output black pixel
    input [10:0] AICrosshairX; //x coordinate of the AI's crosshair
    input [9:0] AICrosshairY; //y coordinate of the AI's crosshair
    input AIFire; //high when AI firing
    input [10:0] playerCrosshairX; //x coordinate of the player's
crosshair
    input [9:0] playerCrosshairY; //y coordinate of the player's
crosshair
    input playerFire; //high when player firing
    input calibrate; //1 to enter calibration mode
    input startPlaying; //1 to start playing
    input leaveCalibrate; //1 to leave calibration mode

    output phsync, pvsync, pblank; //outputted horizontal and vertical
syncs, and blank, with appropriate delay
    output [7:0] pixel; //rgb coloring of pixel
    output [10:0] duckX; //duck's x coordinate
    output [9:0] duckY; //duck's y coordinate
    output duckAlive; //1 if duck is alive
    output [1:0] AIShotsLeft; //number of remaining shots for AI
    output [1:0] playerShotsLeft; //number of remaining shots for player
    output roundOver; //asserted high for one clock cycle if round
expired due to time elapsing
    output roundStart; //high for one clock cycle when round starts
    output [1:0] dir; //direction that duck is travelling
    output inCalibrate; //high when device is in calibration mode

    reg [7:0] pixel;
    reg oldvsync; //stores value of vsync from one clock cycle
ago, used for edge detection

```



```

    reg [10:0] duckX;           //duck's x coordinate
    reg [9:0]  duckY;           //duck's y coordinate
    reg duckAlive;             //1 if duck is alive
    reg [2:0] duckState;       //specifies which image of the duck
should be displayed
    reg [1:0] dir;             //duck's current heading
    reg [3:0] duckSpeed;       //duck's speed

    reg [4:0] playerScore;     //player's score
    reg [4:0] AIScore;         //AI's score
    reg [3:0] roundNumber;     //round number
    reg startTimer;           //asserted to start timer running

    reg signed [4:0] duckXAdj, duckYAdj; //equal to the change in x and y per
frame, signed since x and y could increase or decrease
//these are equal to +/- duckSpeed,
with the sign determined by current direction

    reg [4:0] animCount;       //used to make each frame of duck animation
last for several frames of video output
    reg duckHit; //1 when duck is shot
    reg [1:0] AIShotsLeft; //number of remaining shots for AI
    reg [1:0] playerShotsLeft; //number of remaining shots for player
    reg [7:0] respawnCount; //used to count frames between duck respawn
    reg [19:0] ducksHit; //specifies state of all ducks in round (is alive or
which player killed)
    reg [10:0] popupX; //x coordinate of popup
    reg [9:0]  popupY; //y coordinate of popup
    reg justRespawned; //high for one cycle when the duck just respawned
    reg [10:0] roundX; //x coordinate of top left corner of round display
    reg [9:0]  roundY; //y coordinate of top left corner of round display
    reg [23:0] roundString; //string to display in round display
    reg resetTimer; //asserted to reset timer
    reg playerWin; //1 if player killed duck in last round
    reg oldPlayerFire1; //player fire delayed by one update cycle
    reg oldPlayerFire2;
    reg oldPlayerFire3;
    reg oldPlayerFire4;
    reg oldPlayerFire5;
    reg oldPlayerFire6;

    parameter hDim=1024; //horizontal dimension of screen
    parameter vDim=768; //vertical dimension of screen
    parameter horiz=450; //distance of horizon from top of
screen
    parameter duckWidth=64; //width of duck
    parameter duckHeight=64; //height of duck
    parameter animPeriod=5'b01010; //duration in frames of duck animation
    parameter fallSpeed=4'b0110; //speed of fall

    //enum of directions, north east, north west, and so on
    parameter dirNE = 2'b00;
    parameter dirNW = 2'b01;
    parameter dirSE = 2'b10;
    parameter dirSW = 2'b11;

    //enum of all possible states that the duck could be in, each state
corresponds to a different image
    parameter fly1=3'b000;
    parameter fly2=3'b001;

```

```

parameter fly3=3'b010;
parameter fly4=3'b011;
parameter hit=3'b100;
parameter fall1=3'b101;
parameter fall2=3'b110;
parameter dead=3'b111;

//number of frames before duck respawns
parameter respawnDelay=120; //1 second at 60 hZ refresh rate

//used to specify entries of ducksHit, which holds state of ducks for
score overlay
parameter playerKilledDuck=2'b10;
parameter AIKilledDuck=2'b01;

parameter popupHide=1000;
parameter popupDisplay=338;
parameter borderPixels=30; //number of pixels in border
//connections to game objects
wire [10:0] hCount;
wire [9:0] vCount;
wire [7:0]
backgroundPixel, duckPixel, treePixel, crosshairPixel, AICrosshairPixel, roundPixel,
scorePixel, roundOverPixel, titlePixel;
wire
duckHasPixel, treeHasPixel, crosshairHasPixel, AICrosshairHasPixel, roundOver, score
1, score2, scoreHasPixel, roundOverHasPixel, titleHasPixel, doneTitle, calibrateMode;

//outputted syncs and blank are same as input
assign phsync=hsync;
assign pvsync=vsync;
assign pblank=blank;

assign roundStart=startTimer; //round starts when timer starts
assign inCalibrate=calibrateMode;

always @ (posedge clock) begin
    oldvsync<=vsync;

    if ((reset) || (roundNumber>=10)) begin
        duckX<=hDim/2;
        duckY<=horiz+1;
        dir<=dirNE;
        duckSpeed<=4'b0110;
        duckState<=fly1;
        animCount<=5'b00000;
        duckHit<=0;
        AIShotsLeft<=2'b11;
        playerShotsLeft<=2'b11;
        AIScore<=5'b00000;
        playerScore<=5'b00000;
        duckAlive<=1'b1;
        roundNumber<=4'b0000;
        startTimer<=1'b1;
        respawnCount<=8'b00000000;
        resetTimer<=1'b0;
        ducksHit<=20'b0;
        popupX<=hDim/2;
        popupY<=popupHide; //offscreen
        playerWin<=1'b0;
        oldPlayerFire1<=1'b0;
        oldPlayerFire2<=1'b0;
        oldPlayerFire3<=1'b0;
    end
end

```

```

        oldPlayerFire4<=1'b0;
        oldPlayerFire5<=1'b0;
        oldPlayerFire6<=1'b0;

    end

    //if vsync just stepped low, and title screen is done, update
state
    else if ((oldvsync == 1) && (vsync==0)) && (doneTitle==1)) begin
        oldPlayerFire1<=playerFire;
        oldPlayerFire2<=oldPlayerFire1;
        oldPlayerFire3<=oldPlayerFire2;
        oldPlayerFire4<=oldPlayerFire3;
        oldPlayerFire5<=oldPlayerFire4;
        oldPlayerFire6<=oldPlayerFire5;
        //increment animCount each frame
        if (animCount==animPeriod-1)
            animCount<=5'b00000;
        else
            animCount<=animCount+1;

        //increment animation state
        if (animCount == 5'b00000) begin
            case (duckState)
                fly1: duckState<=duckHit ? hit : (roundOver ?
dead : fly2);
                fly2: duckState<=duckHit ? hit : (roundOver ?
dead : fly3);
                fly3: duckState<=duckHit ? hit : (roundOver ?
dead : fly4);
                fly4: duckState<=duckHit ? hit : (roundOver ?
dead : fly1);
                hit: duckState<=fall1;
                fall1:
duckState<=(duckY+fallSpeed+duckHeight>=horiz) ? dead : fall2;
                fall2:
duckState<=(duckY+fallSpeed+duckHeight>=horiz) ? dead : fall1;
                dead: duckState<=(respawnCount>=respawnDelay)
? fly1 : dead;
                default: duckState<=fly1;
            endcase
        end

        //duck respawning, start of new round
        if(duckState==dead) begin
            duckX<=hDim/2;
            duckY<=horiz-10;
            if(respawnCount>=respawnDelay) begin
                respawnCount<=0;
                AIShotsLeft<=2'b11;
                playerShotsLeft<=2'b11;
                dir<=(roundNumber==((roundNumber/2)*2)) ?

dirNE : dirNW;

                duckAlive<=1'b1;
                duckHit<=0;
                duckState<=fly1; //just in case, already set
above, but done again to avoid any chance of glitch
                startTimer<=1'b1;
                popupY<=popupHide;
            end
            else begin
                popupY<=popupDisplay;

```

```

        respawnCount<=respawnCount+1;
        startTimer<=1'b0;
    end
end
else begin
    startTimer<=1'b0;
end

//determines the Adjustment to duck's x and y coordinates

//if flying, move in given direction
if(duckState==fly1 || duckState==fly2 || duckState==fly3 ||
duckState==fly4) begin
    case (dir)
        dirNE: begin
            duckXAdj<=duckSpeed-
2*(roundNumber-((roundNumber/2)*2));
            duckYAdj<=-duckSpeed;
        end
        dirNW: begin
            duckXAdj<=(-
duckSpeed)+2*(roundNumber-((roundNumber/2)*2));
            duckYAdj<=-duckSpeed;
        end
        dirSE: begin
            duckXAdj<=duckSpeed-
2*(roundNumber-((roundNumber/2)*2));
            duckYAdj<=duckSpeed;
        end
        dirSW: begin
            duckXAdj<=(-
duckSpeed)+2*(roundNumber-((roundNumber/2)*2));
            duckYAdj<=duckSpeed;
        end
    endcase
end

//after being hit
else begin
    //if dead, don't move
    if(duckState==dead) begin
        duckXAdj<=0;
        duckYAdj<=0;
    end

    else begin
        //when first hit don't move
        if(duckState==hit) begin
            duckXAdj<=0;
            duckYAdj<=0;
        end

        //then fall down
        else begin
            duckXAdj<=0;
            duckYAdj<=fallSpeed;
        end
    end
end
end
end

```

```

//if adjustments to x and y coordinate keep the duck in the
sky or behind the grass, just move the duck
    if ((duckX+duckXAdj>=borderPixels) &&
(duckX+duckXAdj+duckWidth<=hDim-borderPixels) && (duckY+duckYAdj>=borderPixels)
&& (duckY+duckYAdj<=horiz) &&(~(duckState==dead))) begin
        duckX<=duckX+duckXAdj;
        duckY<=duckY+duckYAdj;
    end
    else begin
        if(~(duckState==dead)) begin
            case (dir)
                dirNE: dir <= (duckX+duckXAdj+duckWidth
< hDim-borderPixels) ? dirSE : (((duckX+duckXAdj+duckWidth == hDim-
borderPixels) && (duckY+duckYAdj ==borderPixels)) ? dirSW : dirNW);
                dirNW: dir <= (duckX+duckXAdj >
borderPixels) ? dirSW : (((duckX+duckXAdj == borderPixels) && (duckY+duckYAdj
== borderPixels)) ? dirSE : dirNE);
                dirSE: dir <= (duckX+duckXAdj+duckWidth
< hDim-borderPixels) ? dirNE : (((duckX+duckXAdj+duckWidth == hDim-
borderPixels) && (duckY+duckYAdj == horiz)) ? dirNW : dirSW);
                dirSW: dir <= (duckX+duckXAdj >
borderPixels) ? dirNW : (((duckX+duckXAdj == borderPixels) && (duckY+duckYAdj
== horiz)) ? dirNE : dirSE);
            endcase

            //moves the duck during reflection
            duckX<=((duckX+duckXAdj+duckWidth < hDim-
borderPixels) && (duckX+duckXAdj>borderPixels)) ? duckX+duckXAdj : duckX-
duckXAdj;

            duckY<=((duckY+duckYAdj < horiz) &&
(duckY+duckYAdj>borderPixels)) ? duckY+duckYAdj : duckY-duckYAdj;
        end
    end

//handles player shooting
    if(oldPlayerFire5 && ~(oldPlayerFire6)) &&
(playerShotsLeft>0)) begin
        playerShotsLeft<=playerShotsLeft-1;

        if((playerCrosshairX>=duckX)&&(playerCrosshairX<duckX+duckWidth)&&(player
CrosshairY>=duckY)&&(playerCrosshairY<duckY+duckHeight)) begin
            duckHit<=1;
            duckAlive<=1'b0;
            roundNumber<=roundNumber+1;
            resetTimer<=1'b1;
            playerWin<=1'b1;
            if(score2) begin
                playerScore<=playerScore+1;
            end
            else begin
                if(score1) playerScore<=playerScore+2;
                else playerScore<=playerScore+3;
            end
            end
            case (roundNumber)
                0:ducksHit[1:0]<=playerKilledDuck;
                1:ducksHit[3:2]<=playerKilledDuck;
                2:ducksHit[5:4]<=playerKilledDuck;
                3:ducksHit[7:6]<=playerKilledDuck;
                4:ducksHit[9:8]<=playerKilledDuck;
                5:ducksHit[11:10]<=playerKilledDuck;
                6:ducksHit[13:12]<=playerKilledDuck;
                7:ducksHit[15:14]<=playerKilledDuck;
                8:ducksHit[17:16]<=playerKilledDuck;
            endcase
        end
    end
end

```



```

TreeGrass aTree(clock,hCount,vCount,treeHasPixel, treePixel);

//player's crosshair
Crosshair aCrosshair(clock, playerCrosshairX, playerCrosshairY, hCount,
vCount, crosshairHasPixel, crosshairPixel);
defparam aCrosshair.color = 8'b00011100;

//ai's crosshair
Crosshair bCrosshair(clock, AICrosshairX, AICrosshairY, hCount, vCount,
AICrosshairHasPixel, AICrosshairPixel);
defparam bCrosshair.color = 8'b11100000;

//round timer
RoundTimer aRoundTimer(clock,reset,startTimer,resetTimer,roundOver);
defparam aRoundTimer.numSeconds = 10;

//one second score timer
RoundTimer scoreTimer1(clock,reset,startTimer,resetTimer,score1);
defparam scoreTimer1.numSeconds = 1;

//three second score timer
RoundTimer scoreTimer2(clock,reset,startTimer,resetTimer,score2);
defparam scoreTimer2.numSeconds = 3;

//score overlay
ScoreOverlay
aScoreOverlay(clock,reset,hCount,vCount,playerScore,playerShotsLeft,ducksHit,sc
oreHasPixel,scorePixel);

//round over overlay
RoundOverOverlay
aRoundOverOverlay(clock,reset,roundOver,hCount,vCount,popupX,popupY,playerWin,r
oundOverHasPixel,roundOverPixel);

wire compoundReset;
assign compoundReset=(roundNumber>=10);
//title screen
TitleScreen
aTitleScreen(clock,reset,compoundReset,hCount,vCount,calibrate,startPlaying,lea
veCalibrate,titleHasPixel,titlePixel,doneTitle,calibrateMode);
//round display font rom, still need to implement roundX,roundY, and
roundString calculations
//char_string_display
aChar_string_display(clock,hcount,vcount,roundString,roundX,roundY,roundPixel);
//defparam aChar_string_display.NCHAR=4;
//defparam aChar_string_display.NCHAR_BITS=2;

//update pixel whenever a new location is requested, will be cleaned up
to avoid all this if/else
always @ * begin
if(doneTitle) begin
if (scoreHasPixel) pixel<=scorePixel;
else begin
if (crosshairHasPixel) pixel<=crosshairPixel;
else begin
if (AICrosshairHasPixel)
pixel<=AICrosshairPixel;
else begin
if (treeHasPixel)
pixel<=treePixel;
else begin

```

```

pixel<=roundOverPixel;

pixel<=duckPixel;

pixel<=backgroundPixel;

                                     if (roundOverHasPixel)
                                     else begin
                                         if (duckHasPixel)
                                         else
                                             end
                                         end
                                     end
                                     end
                                     end
                                     end
                                         end
                                         end
                                         end
                                             end
                                             end
                                             end
                                                 pixel<=titlePixel;
                                             end
                                                 end
                                                 end
                                                     end
                                                     end
                                                     end
                                                         endmodule

```

Graphics Pipeline

Background Sky

```

////////////////////////////////////////////////////
//
// Background:blue sky
//
////////////////////////////////////////////////////
module Background (hCount,vCount,pixel);
    input [10:0] hCount;          //hCount is x coordinate of current pixel
being requested
    input [9:0] vCount;          //vCount is y coordinate of the pixel being
requested

    output [7:0] pixel; //rgb value of pixel

    parameter hDim=1024; //hDim is horizontal size of screen
    parameter vDim=768; //vDim is vertical size of screen
    parameter horizon=450; //location of horizon as measured from top of
screen
    parameter topColor=8'b00000111; //color of sky
    parameter bottomColor=8'b11111010; //color of dirt, never actually
displayed, grass and dirt display over it, there just in case
    parameter blackColor=8'b00000000; //border
    parameter dBlack=30; //30 pixel black border
    assign pixel= ((hCount<=dBlack) || (hCount>hDim-dBlack) ||
(vCount<=dBlack)) ? blackColor : (((vCount >= 0) && (vCount <= horizon+50)) ?
topColor : bottomColor);

endmodule

```

Duck

```

////////////////////////////////////////////////////
//
// Duck: animated duck
//
////////////////////////////////////////////////////

```



```

module Duck (clk, x, y, hCount, vCount, state, hasPixel, pixel);
    input clk; //clock
    input [10:0] x, hCount; //x is the duck's x position, hCount is the x
coordinate of the pixel being requested
    input [9:0] y, vCount; //y is the duck's y position, vCount is
the y coordinate of the pixel being requested
    input [2:0] state; //specifies which of the possible images of
the duck to display, like fly or falling

    output hasPixel; //asserted high if the duck has a pixel in
that region
    output [7:0] pixel; //rgb value of the pixel

    reg hasPixel; //asserted high if the duck has a pixel in
that region
    reg [7:0] pixel; //rgb value of the pixel

    reg [11:0] addr; //addr to read from ram

    wire [7:0] dout1,dout2,dout3,dout4,dout5,dout6;
//enum of all possible states that the duck could be in, each state
corresponds to a different image
    parameter fly1=3'b000;
    parameter fly2=3'b001;
    parameter fly3=3'b010;
    parameter fly4=3'b011; //same as fly2 in terms of image
    parameter hit=3'b100;
    parameter fall1=3'b101;
    parameter fall2=3'b110;
    parameter dead=3'b111;

    parameter flyWidth=64;
    parameter fallWidth=32;
    parameter height=64;

    wire [6:0] width;

//these roms contain the bitmaps of the duck in various positions
    fly1rom aFly1Rom(addr, clk, dout1);
    fly2rom aFly2Rom(addr, clk, dout2);
    fly3rom aFly3Rom(addr, clk, dout3);
    hitrom aHitRom(addr, clk, dout4);
    fall1rom aFall1Rom(addr, clk, dout5);
    fall2rom aFall2Rom(addr, clk, dout6);

//width of duck sprite different when flying or falling
    assign width=(state==fly1 || state==fly2 || state==fly3 || state==fly4 ||
state==hit) ? flyWidth : fallWidth;
    always @ (posedge clk) begin
        //check to see if requested pixel is within boundary
        if ((hCount >= x) && (hCount < x+width) && (vCount >= y) &&
(vCount < y+height)) begin
            hasPixel<=1;
            //determine which address to read
            addr<=width*(vCount-y)+(hCount-x);

            //determine which output to use
            case (state)
                fly1:pixel<=dout1;
                fly2:pixel<=dout2;
                fly3:pixel<=dout3;
                fly4:pixel<=dout2;
            endcase
        end
    end
endmodule

```

```

        hit:pixel<=dout4;
        fall1:pixel<=dout5;
        fall2:pixel<=dout6;
        default:pixel<=dout1;
    endcase
end
    else begin
        hasPixel<=0;
    end
end
endmodule

```

Tree and Grass

```

////////////////////////////////////
//
// TreeGrass: tree and grass layer
//
////////////////////////////////////
module TreeGrass (clk, hCount, vCount, hasPixel, pixel);
    input clk; //clock
    input [10:0] hCount; //horizontal location of the
pixel being requested
    input [9:0] vCount; //vertical location of the pixel being
requested

    output hasPixel; //asserted high if this object
has a pixel at hCount, vCount
    output [7:0] pixel; //rgb value of pixel

    reg hasPixel;
    reg [7:0] pixel;

    reg [12:0] addr1;
    reg [14:0] addr2;

    wire [7:0] dout1,dout2;

    wire [10:0] hCountPrime;
    wire [9:0] vCountPrime;

    parameter height=460;
    parameter width=276;
    parameter x=100; //x coordinate of top left corner of tree
    parameter y=30; //y coordinate of top left corner of tree

    parameter horizon=450; //location of horizon
    parameter grassDirtHeight=360; //height of grass and dirt
    parameter screenHeight=768;
    parameter screenWidth=1024;
    parameter hBlack=30; //leftmost and rightmost hBlack pixels are
black in grass and dirt section

    assign hCountPrime=hCount/4;
    assign vCountPrime=vCount/4;

    //roms that store tree, grass, dirt
    treerom aTreeRom(addr1,clk,dout1);
    grassdirtrom aGrassDirtRom(addr2,clk,dout2);

    always @ (posedge clk) begin

```

```

        //check if pixel is within boundary of grass, divisions by 4 are
due to compression of grass bitmap
        if ((vCount>(screenHeight-grassDirtHeight)) &&
(vCount<=screenHeight)) begin
            if((hCount<hBlack) || (hCount>(screenWidth-hBlack))) begin
                pixel<=8'b00000000;
                hasPixel<=1'b1;
            end
            else begin
                addr2<=(screenWidth/4)*(vCountPrime-(screenHeight-
grassDirtHeight)/4)+hCountPrime; //bitmap for grass and dirt is compressed by 4
                if(dout2==8'b0000111) begin //if pixel is blue,
don't display so duck will be on top of blue sky but behind grass
                    //check to see if requested pixel is within
boundary of tree, division by 4 due to compression of tree bitmap
                    if ((hCount >= x) && (hCount < x+width) &&
(vCount >= y) && (vCount < y+height)) begin
                        //determine which address to read
                        addr1<=width/4*(vCount/4-
y/4)+(hCount/4-x/4);

                        pixel<=dout1;
                        hasPixel<=1'b1;
                    end
                    else
                        hasPixel<=1'b0;
                    end
                else begin
                    pixel<=dout2;
                    hasPixel<=1'b1;
                end
            end
        end
    end
    else begin
        //check to see if requested pixel is within boundary of
tree, division by 4 due to compression of tree bitmap
        if ((hCount >= x) && (hCount < x+width) && (vCount >= y) &&
(vCount < y+height)) begin
            //determine which address to read
            addr1<=width/4*(vCount/4-y/4)+(hCount/4-x/4);

            pixel<=dout1;
            hasPixel<=~(dout1==8'b0000111); //if pixel
is blue, don't display so duck will be on top of blue sky but behind tree
        end
        else begin
            hasPixel<=1'b0;
        end
    end
end
end

endmodule

```

Crosshair

```

////////////////////////////////////
//

```

```

// Crosshair
//
////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////
module Crosshair (clk, x, y, hCount, vCount, hasPixel, pixel);
    input clk; //clock
    input [10:0] x; //x coordinate of
crosshair(center)
    input [9:0] y; //y coordinate of
crosshair(center)
    input [10:0] hCount; //horizontal location of the
pixel being requested
    input [9:0] vCount; //vertical location of the pixel being
requested

    output hasPixel; //asserted high if this object
has a pixel at hCount, vCount
    output [7:0] pixel; //rgb value of pixel

    reg hasPixel;
    reg [7:0] pixel;
    reg [7:0] addr;
    wire dout;

    parameter height=16; //height of crosshair
    parameter width=16; //width of crosshair
    parameter color=8'b11111111; //color of crosshair
    parameter hOff=0; //memory offset to deal with delay

    crosshairrom aCrosshairRom(addr,clk,dout);

    always @ (posedge clk) begin
        //check to see if requested pixel is within boundary, horiz offset
of 1 for timing reasons
        if ((hCount+hOff >= x-(width/2)) && (hCount+hOff < x+(width/2)) &&
(vCount >= y-(height/2)) && (vCount < (y+height/2))) begin
            //determine which address to read
            addr<=width*(vCount-(y-height/2))+(hCount+hOff-(x-
width/2));

            hasPixel<=dout;
            pixel<=color;
        end
        else begin
            hasPixel<=0;
        end
    end
endmodule

```

AI

```

////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////
//
// AI
//
////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////
module
AI (clk, reset, curDuckX, curDuckY, duckAlive, shotsLeft, roundStart, roundTimeExpired,
dir, vsync, difficulty, x, y, fire);
    input clk; //clock
    input reset; //global reset, active high

```

```

input [10:0] curDuckX; //duck's x coordinate
input [9:0] curDuckY; //duck's y coordinate
input duckAlive; //1 if duck is alive
input [1:0] shotsLeft; //number of shots left this round
input roundStart; //high for one clock cycle at start of round
input roundTimeExpired; //high if round timer expires
input [1:0] dir; //direction that duck is travelling
input vsync; //vertical sync signal
input [1:0] difficulty; //AI difficult, 00 is very hard, 01 is medium, 10
is easy, 11 is very easy

output [10:0] x; //x coordinate of crosshair
output [9:0] y; //y coordinate of crosshair
output fire; //high for one cycle when firing

reg [10:0] x; //x coordinate of crosshair
reg [9:0] y; //y coordinate of crosshair
reg fire; //high for one cycle when firing
reg justFired; //high for ten cycles after firing
reg [3:0] fireTimeCount; //used to count time since last firing
reg freeToShoot; //1 when AI is allowed to shoot
reg roundOver; //1 if round is over
reg resetTimer; //asserted to reset timer
reg [10:0] oldDuckX; //value of curDuckX from last clock cycle
reg [9:0] oldDuckY; //value of curDuckY from last clock cycle
reg [10:0] duckX; //estimated value of new curDuckX
reg [9:0] duckY; //estimated value of new curDuckY

parameter hDim=1024; //horizontal dimension of screen
parameter vDim=768; //vertical dimension of screen

parameter duckWidth=64; //width of duck
parameter duckHeight=64; //height of duck

parameter maxMove=40; //largest distance that the crosshair can move per
update, in each of x or y
parameter shotAdj=8+(4*difficulty); //min distance from border to
take shot
parameter shotDelay=7; //number of seconds between start of round
before attempting shooting

//enum of directions, north east, north west, and so on
parameter dirNE = 2'b00;
parameter dirNW = 2'b01;
parameter dirSE = 2'b10;
parameter dirSW = 2'b11;

wire delayOver;
//round timer
RoundTimer shotDelayTimer(clk,reset,roundStart,resetTimer,delayOver);
defparam shotDelayTimer.numSeconds = shotDelay+difficulty;

always @ (posedge clk) begin
roundOver<=(~duckAlive) || roundTimeExpired);
oldDuckX<=curDuckX;
oldDuckY<=curDuckY;
duckX<=(2*curDuckX)-oldDuckX;
duckY<=(2*curDuckY)-oldDuckY;
resetTimer<=1'b0;

```

```

if (reset) begin
    x<=hDim/2;
    y<=vDim/4;
    justFired<=0;
    fireTimeCount<=4'b0000;
    fire<=1'b0;
    freeToShoot<=1'b0;
    oldDuckX<=curDuckX;
    oldDuckY<=curDuckY;
end
else begin

    //determines when AI is free to shoot
    if(roundOver) begin
        freeToShoot<=1'b0;
    end
    else begin
        if(delayOver) begin
            freeToShoot<=1'b1;
        end
    end

    //if duck just moved
    if(~((curDuckX==oldDuckX) && (curDuckY==oldDuckY))) begin
        //if x coordinate near duck x center, only move a
        little
        if ((x>=duckX+duckWidth/2-maxMove) &&
(x<duckX+duckWidth/2+maxMove)) begin
            x<=duckX+duckWidth/2;
        end
        //otherwise move a lot
        else begin
            if (x<duckX+duckWidth/2-maxMove) x<=x+maxMove;
            else x<=x-maxMove;
        end

        //if y coordinate near duck y center, only move a
        little
        if ((y>=duckY+duckHeight/2-maxMove) &&
(y<duckY+duckHeight/2+maxMove)) begin
            y<=duckY+duckHeight/2;
        end
        //otherwise move a lot
        else begin
            if (y<duckY+duckHeight/2-maxMove)
y<=y+maxMove;
            else y<=y-maxMove;
        end

        //handles firing
        if ((x>=duckX+shotAdj) && (x<duckX+duckWidth-shotAdj)
&& (y>=duckY+shotAdj) && (y<duckY+duckHeight-shotAdj) && (shotsLeft>0) &&
(freeToShoot==1'b1) && duckAlive && ~justFired) begin
            justFired<=1;
            fireTimeCount<=0;
            fire<=1'b1;
        end
        else fire<=1'b0;
        //increment fireTimeCount and reset justFired to 0
        when fireTimeCount reaches 10
        if (justFired) begin

```

```

        if (fireTimeCount < 10)
fireTimeCount <= fireTimeCount + 1;
        else justFired <= 0;
        end
    end
end
end
end

endmodule

////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////
//
// playerInput (temporary, for test purposes only)
//
////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////
module
playerInput (clk, reset, moveLeft, moveRight, moveUp, moveDown, fireIn, shotsLeft, x, y, fireOut);
    input clk, reset, moveLeft, moveRight, moveUp, moveDown, fireIn;
    input [1:0] shotsLeft;

    output [10:0] x;
    output [9:0] y;
    output fireOut;

    reg [10:0] x;
    reg [9:0] y;

    parameter screenWidth=1024;
    parameter screenHeight=768;
    parameter posAdj=20;

    reg oldMoveLeft, oldMoveRight, oldMoveUp, oldMoveDown, oldFireIn, fireOut;
    reg [1:0] oldShotsLeft;

    always @ (posedge clk) begin
        oldMoveLeft <= moveLeft;
        oldMoveRight <= moveRight;
        oldMoveUp <= moveUp;
        oldMoveDown <= moveDown;
        oldFireIn <= fireIn;
        oldShotsLeft <= shotsLeft;
        if (reset) begin
            x <= screenWidth/2;
            y <= screenHeight/2;
        end
        else begin
            if (moveLeft && !oldMoveLeft) begin
                x <= x - posAdj;
            end
            else begin
                if (moveRight && !oldMoveRight) begin
                    x <= x + posAdj;
                end
            end

            if (moveUp && !oldMoveUp) begin
                y <= y - posAdj;
            end
            else begin
                if (moveDown && !oldMoveDown) begin

```

```

        y<= y+posAdj;
    end
end
if((fireIn==1) && (oldFireIn==0)) begin
    fireOut<=1;
end
else begin
    if((fireOut==1) && ~(oldShotsLeft==shotsLeft))
        fireOut<=0;
    end
end
end
endmodule

```

Round Timer

```

////////////////////////////////////
//
// RoundTimer: timer to determine when round ends
//
////////////////////////////////////
module RoundTimer(clk,reset,startTimer,resetTimer,timeElapsed);
    input clk; //clock
    input reset; //global reset
    input startTimer; //high for 1 cycle when the timer should be started
    input resetTimer; //if asserted, resets timer to 0;
    output timeElapsed; //high for 1 cycle when time elapsed

    reg timeElapsed; //high when time elapsed
until next startTimer
    reg currentlyTiming; //1 when timing
    reg [0:29] count; //used to count clock
cycles, excessively large to allow larger

    //timing intervals to be implemented without danger of overflow
parameter numSeconds=5; //number of seconds before round over
parameter clockFreq=65000000; //clock frequency

always @ (posedge clk) begin
    if(reset || resetTimer) begin
        count<=30'b0;
        currentlyTiming<=1'b0;
        timeElapsed<=1'b0;
    end
    else begin
        if(startTimer && !currentlyTiming) begin
            currentlyTiming<=1'b1;
            timeElapsed<=1'b0;
        end
        else begin
            if(currentlyTiming && (count>=numSeconds*clockFreq))
begin
                timeElapsed<=1'b1;
                count<=30'b0;
                currentlyTiming<=1'b0;
            end
            else begin
                count<=count+1;
                //timeElapsed<=1'b0;
            end
        end
    end
end

```



```

        end
    end
endmodule

```

Score Overlay

```

////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////
//
// Score overlay
//
////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////////
module
ScoreOverlay(clk,reset,hCount,vCount,playerScore,playerShotsLeft,ducksHit,hasPi
xel,pixel);
    input clk; //system clock
    input reset; //global reset
    input [10:0] hCount; //horizontal location of the
pixel being requested
    input [9:0] vCount; //vertical location of the pixel being
requested
    input [4:0] playerScore; //players score in hundreds
    input [1:0] playerShotsLeft; //number of shots remaining
    input [19:0] ducksHit; //which ducks were hit by whom, 10
pairs, 00 means not hit, 10 means hit by player, 01 means hit by AI

    output hasPixel; //asserted high if this object
has a pixel at hCount, vCount
    output [7:0] pixel; //rgb value of pixel

    reg [7:0] addr1,addr2,addr3;
    reg [3:0] selectedDuck; //duck whose pixel is currently
being requested
    reg [1:0] selectedDuckState; //state of selected duck
    reg [1:0] selectedBullet;

    reg hasPixel;
    reg [7:0] pixel;

    reg hasPixel1,hasPixel2,hasPixel3,pixel1;
    reg [7:0] fpixel1,pixel2,pixel3;
    reg [3:0] digit1, digit2; //first and second digit of score

    wire [7:0] doutB;
    wire doutA,dout0,dout1,dout2,dout3,dout4,dout5,dout6,dout7,dout8,dout9;

    parameter deadDuckX=360; //x coordinate of top left corner of
first duck
    parameter deadDuckY=628; //y coordinate of top left corner of
first duck
    parameter duckHeight=16; //height of duck
    parameter duckWidth=16; //width of duck
    parameter numDucks=10; //number of ducks per round
    parameter bulletX=112; //x coordinate of top left corner of first
bullet
    parameter bulletY=628; //y coordinate of top left corner of first
bullet
    parameter bulletWidth=16; //width of bullet
    parameter bulletHeight=15; //height of bullet

```

```

    parameter scoreX=812;          //x coordinate of top left corner of first
score digit
    parameter scoreY=612;          //y coordinate of top left corner of
first score digit
    parameter scoreWidth=16;      //width of score digit
    parameter scoreHeight=16;     //height of score digit
    parameter scoreDigits=4;      //number of digits in score

//enum of duck state and color
parameter duckAlive=2'b00;
parameter duckDeadPlayer=2'b10;
parameter duckDeadAI=2'b01;
parameter aliveColor=8'b11111111;
parameter playerColor=8'b00011100;
parameter AIColor=8'b11100000;

scoreduckrom aScoreDuckRom(addr2,clk,doutA);
scorebulletrom aScoreBulletRom(addr3,clk,doutB);
score0rom aScore0rom(addr1,clk,dout0);
score1rom aScore1rom(addr1,clk,dout1);
score2rom aScore2rom(addr1,clk,dout2);
score3rom aScore3rom(addr1,clk,dout3);
score4rom aScore4rom(addr1,clk,dout4);
score5rom aScore5rom(addr1,clk,dout5);
score6rom aScore6rom(addr1,clk,dout6);
score7rom aScore7rom(addr1,clk,dout7);
score8rom aScore8rom(addr1,clk,dout8);
score9rom aScore9rom(addr1,clk,dout9);

always @ (posedge clk) begin
    if(reset) begin
        addr2<=8'b0;
        selectedDuck<=4'b0;
    end
    else begin

        //compute score digits
        if(playerScore<10)
            digit1<=0;
        else begin
            if(playerScore<20)
                digit1<=1;
            else begin
                if(playerScore<30)
                    digit1<=2;
                else
                    digit1<=3;
            end
        end
        end
        digit2<=playerScore-digit1*10;
        //check if requested pixel is within boundary of any duck
        if((hCount>=deadDuckX) &&
(hCount<deadDuckX+numDucks*duckWidth) && (vCount>=deadDuckY) &&
(vCount<deadDuckY+duckHeight)) begin
            selectedDuck<=(hCount-deadDuckX)/duckWidth;
            case (selectedDuck)
                0: selectedDuckState<=ducksHit[1:0];
                1: selectedDuckState<=ducksHit[3:2];
                2: selectedDuckState<=ducksHit[5:4];
                3: selectedDuckState<=ducksHit[7:6];
            end
        end
    end
end

```

```

        4: selectedDuckState<=ducksHit[9:8];
        5: selectedDuckState<=ducksHit[11:10];
        6: selectedDuckState<=ducksHit[13:12];
        7: selectedDuckState<=ducksHit[15:14];
        8: selectedDuckState<=ducksHit[17:16];
        9: selectedDuckState<=ducksHit[19:18];
        default: selectedDuckState<=ducksHit[1:0];
    endcase
    addr2<=hCount-deadDuckX-
(selectedDuck*duckWidth)+(vCount-deadDuckY)*duckHeight;
    hasPixel2<=doutA;
    case (selectedDuckState)
        duckAlive: pixel2<=aliveColor;
        duckDeadPlayer: pixel2<=playerColor;
        duckDeadAI: pixel2<=AIColor;
        default: pixel2<=aliveColor;
    endcase
end
else begin
    pixel2<=8'b0;
    hasPixel2<=1'b0;
end

//check if requested pixel is within boundary of bullets
if((hCount>=bulletX) &&
(hCount<bulletX+bulletWidth*playerShotsLeft) && (vCount>=bulletY) &&
(vCount<bulletY+bulletHeight)) begin
    hasPixel3<=1'b1;
    pixel3<=doutB;
    selectedBullet<=(hCount-bulletX)/(bulletWidth);
    addr3<=hCount-bulletX-
(selectedBullet*bulletWidth)+(vCount-bulletY)*bulletHeight;
end
else begin
    pixel3<=8'b0;
    hasPixel3<=1'b0;
end

//check if requested pixel is within boundary of score
if((hCount>=scoreX) &&
(hCount<scoreX+scoreWidth*2*scoreDigits) && (vCount>=scoreY) &&
(vCount<scoreY+scoreHeight*2)) begin
    //first digit
    if(hCount<scoreX+scoreWidth*2) begin
        hasPixel1<=1'b1;
        addr1<=(hCount-scoreX)/2+((vCount-
scoreY)/2)*(scoreHeight);
        case (digit1)
            0:pixel1<=dout0;
            1:pixel1<=dout1;
            2:pixel1<=dout2;
            3:pixel1<=dout3;
            4:pixel1<=dout4;
            5:pixel1<=dout5;
            6:pixel1<=dout6;
            7:pixel1<=dout7;
            8:pixel1<=dout8;
            9:pixel1<=dout9;
            default:pixel1<=dout0;
        endcase
    end
    else begin
        //second digit

```

```

        if(hCount<scoreX+scoreWidth*4) begin
            hasPixel1<=1'b1;
            addr1<=(hCount-scoreX-
scoreWidth*2)/2+((vCount-scoreY)/2)*(scoreHeight);
            case (digit2)
                0:pixel1<=dout0;
                1:pixel1<=dout1;
                2:pixel1<=dout2;
                3:pixel1<=dout3;
                4:pixel1<=dout4;
                5:pixel1<=dout5;
                6:pixel1<=dout6;
                7:pixel1<=dout7;
                8:pixel1<=dout8;
                9:pixel1<=dout9;
                default:pixel1<=dout0;
            endcase
        end
        else begin
            //third digit
            if(hCount<scoreX+scoreWidth*6) begin
                hasPixel1<=1'b1;
                pixel1<=dout0;
                addr1<=(hCount-scoreX-
scoreWidth*4)/2+((vCount-scoreY)/2)*(scoreHeight);
            end
            //fourth digit
            else begin
                hasPixel1<=1'b1;
                pixel1<=dout0;
                addr1<=(hCount-scoreX-
scoreWidth*6)/2+((vCount-scoreY)/2)*(scoreHeight);
            end
        end
        end
        end
        else begin
            pixel1<=8'b0;
            hasPixel1<=1'b0;
        end

        //determine output
        fpixel1<=(pixel1==1'b1)? 8'b11111111: 8'b00000000;
        pixel<=fpixel1|pixel2|pixel3;
        hasPixel<=hasPixel1|hasPixel2|hasPixel3;

    end
end
endmodule

```

Round Over Overlay

```

////////////////////////////////////
//
// Round over overlay
//
////////////////////////////////////
module
RoundOverOverlay(clk,reset,roundOver,hCount,vCount,x,y,win,hasPixel,pixel);
    input clk;    //system clock
    input reset; //global reset
    input roundOver;    //asserted high when round over

```

```

        input [10:0] hCount;                //horizontal location of the
pixel being requested
        input [9:0] vCount;                //vertical location of the pixel being
requested
        input [10:0] x; //x coordinate of top left corner of cheney popup
        input [9:0] y; //y coordinate of top left corner of cheney popup
        input win; //1 if player killed duck this round

        output hasPixel;                  //asserted high if this object
has a pixel at hCount, vCount
        output [7:0] pixel;                //rgb value of pixel

        reg hasPixel;
        reg [7:0] pixel;

        parameter popWidth=128;
        parameter popHeight=96;

        reg [13:0] addr;
        wire [7:0] dout,dout2;

        popuprom aPopupRom(addr,clk,dout);
        popuprom2 aPopupRom2(addr,clk,dout2);
        always @ (posedge clk) begin
            if((hCount>=x) && (hCount<x+popWidth) && (vCount>=y) &&
(vCount<y+popHeight)) begin
                hasPixel<=1'b1;
                pixel<=(win ? dout2: dout);
                addr<=hCount-x+(vCount-y)*popWidth;
            end
            else begin
                hasPixel<=1'b0;
            end
        end
    end
endmodule

```

Title Screen

```

////////////////////////////////////
//
// title screen
//
////////////////////////////////////
module
TitleScreen(clk,reset,compoundReset,hCount,vCount,calibrate,startPlaying,leaveC
alibrate,hasPixel,pixel,done,calibrateMode);
    input clk; //system clock
    input reset; //global reset
    input compoundReset; //reset due to restarting round
    input [10:0] hCount; //horizontal location of the
pixel being requested
    input [9:0] vCount; //vertical location of the pixel being
requested
    input calibrate; //asserted to enter calibration
mode
    input startPlaying; //asserted to start playing
    input leaveCalibrate; //asserted to leave calibrate

    output hasPixel; //asserted high if this object
has a pixel at hCount, vCount
    output [7:0] pixel; //rgb value of pixel

```

```

output done;                                //high when module finished
output calibrateMode;                       //high when in calibration mode

reg hasPixel,done;
reg [7:0] pixel;

reg [13:0] addr;
wire [7:0] dout;

reg calibrateMode, startPlayingMode;
titlerom aTitleRom(addr,clk,dout);

parameter screenWidth=1024;
parameter screenHeight=768;
parameter hBlack=8; //first and last 16 pixels are black

always @ (posedge clk) begin
    if(reset || compoundReset) begin
        calibrateMode<=1'b0;
        startPlayingMode<=1'b0;
        done<=1'b0;
    end
    //latch calibrate and startPlaying
    if(calibrate) calibrateMode<=1'b1;
    if(startPlaying) startPlayingMode<=1'b1;

    if(calibrateMode) begin
        done<=1'b0;
        calibrateMode<=leaveCalibrate ? 1'b0 :1'b1;
    end
    else begin
        if(startPlayingMode) begin
            done<=1'b1;
        end
        else begin
            hasPixel<=1'b1;
            pixel<=((hCount<=hBlack) || (hCount>=screenWidth-
hBlack)) ? 8'b0:dout;
            //divisions by 8 are due to compression of title
            bitmap
            addr<=hCount/8+((vCount/8)*(screenWidth/8));
            done<=1'b0;
        end
    end
end
end
endmodule

```

```
////////////////////////////////////  
// parameterized delay line  
  
module delayN(clk,in,out);  
    input clk;  
    input in;  
    output out;  
  
    parameter NDELAY = 3;  
  
    reg [NDELAY-1:0] shiftreg;  
    wire          out = shiftreg[NDELAY-1];  
  
    always @(posedge clk)  
        shiftreg <= {shiftreg[NDELAY-2:0],in};  
  
endmodule // delayN
```